

Lecture – 17 and 18

Date: 23.03.2015

- Inductors and Inductances
- Magnetic Energy
- Magnetic Circuits
- Electromagnetic Fields

Inductors

- Generally - coil of conducting wire
 - Usually wrapped around a solid core. If no core is used, then the inductor is said to have an 'air core'.

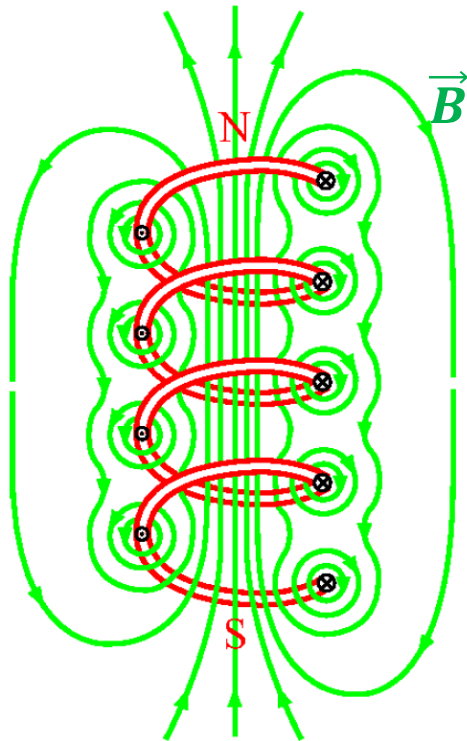


- An inductor is the magnetic analogue of an electric capacitor.
- Just as a capacitor can store energy in the electric field in the medium between its conducting surfaces, an inductor can store energy in the magnetic field near its current carrying conductors.

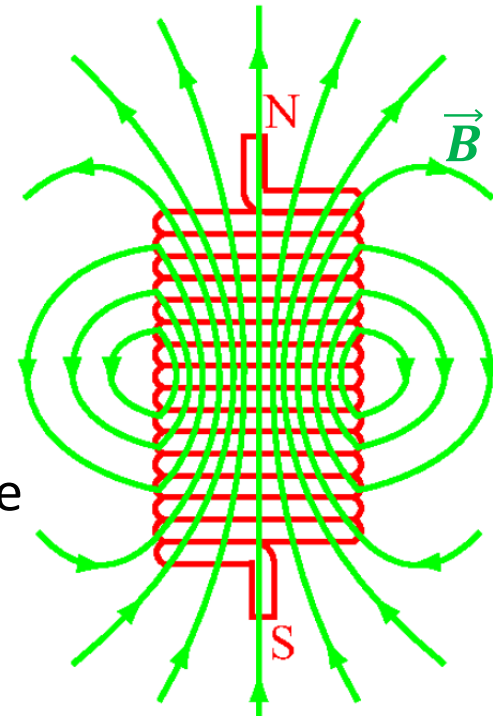
Inductors (contd.)

- A typical inductor consists of multiple turns of wire helically coiled around a cylindrical core called a solenoid.

- Core may be air-filled or may contain a magnetic material with permeability μ .
- If the turns are closely spaced, the solenoid will create better inductor.

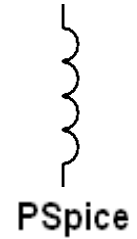
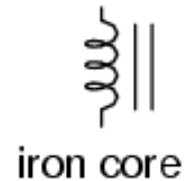
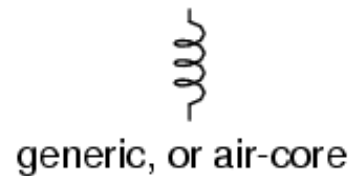


The magnetic field lines resembles those of the permanent magnet



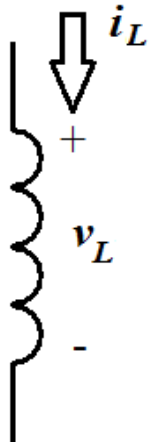
Inductors (contd.)

- Symbols



- Sign Convention

- The sign convention used with an inductor is the same as for a power dissipating device.
 - When current flows into the positive side of the voltage across the inductor, it is positive and the inductor is dissipating power.
 - When the inductor releases energy back into the circuit, the sign of the current will be negative.



Inductors (contd.)

- In such architecture, the current is proportional to the flux linkage:

$$\lambda = LI$$

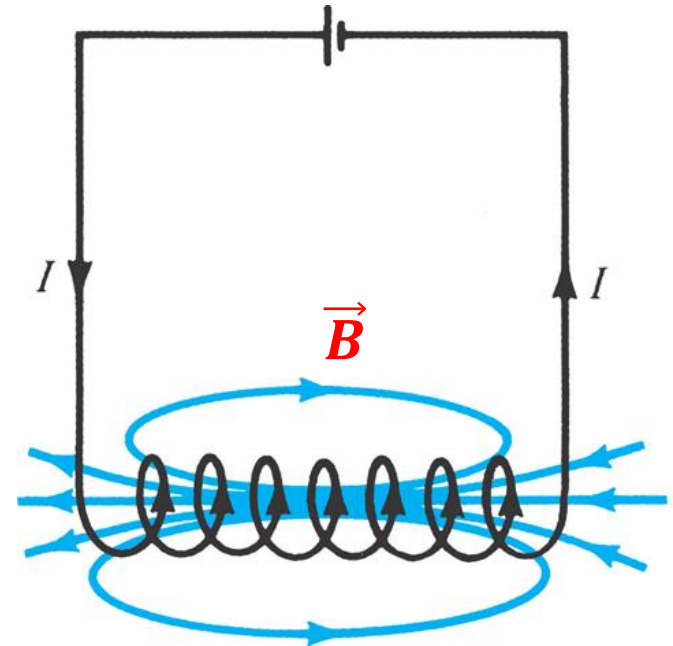
where, L is a constant of proportionality called the **inductance** of the circuit.

- A circuit (or closed conducting path) carrying current I produces a magnetic field \vec{B} that causes a flux $\psi = \int \vec{B} \cdot \vec{ds}$ to pass through each turn of the circuit. Circuit with N identical turns has flux linkage of:
- Inductance, L , is then defined as the ratio of the magnetic flux linkage λ to the current I through the inductor as:

$$\lambda = N\psi$$

$$L = \frac{\lambda}{I} = \frac{N\psi}{I}$$

The unit is Henry (i.e, Wb/A).



Inductors (contd.)

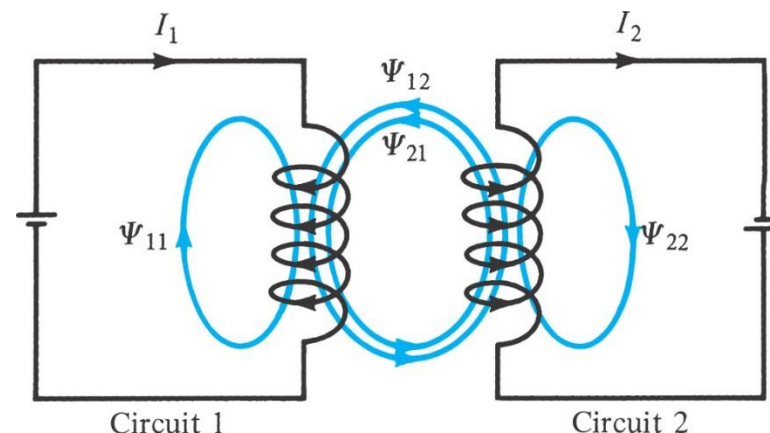
- Like capacitance, inductance may also be regarded as a measure of how much magnetic energy is stored in an inductor.
- The magnetic energy stored in an inductor is:

$$W_m = \frac{1}{2} LI^2$$



$$L = \frac{2W_m}{I^2}$$

- In case of two circuits carry current I_1 and I_2 as shown then a magnetic interaction exists between the circuits.



- Four component fluxes ψ_{11} , ψ_{12} , ψ_{21} , and ψ_{22} are produced. ψ_{12} , for example, is the flux passing through circuit-1 due to current I_2 in circuit-2.
- If \vec{B}_2 is the field due to I_2 and S_1 is the area of circuit-1 then:

$$\Psi_{12} = \int_{S_1} \vec{B}_2 \cdot \vec{ds}$$

Inductors (contd.)

- The mutual inductance M_{12} is defined as the ratio of the flux linkage $\lambda_{12} = N_1\psi_{12}$ of circuit-1 to current I_2 :

$$M_{12} = \frac{\lambda_{12}}{I_2} = \frac{N_1\Psi_{12}}{I_2}$$

- Similarly, M_{21} is defined as the ratio of the flux linkage $\lambda_{21} = N_2\psi_{21}$ of circuit-2 to current I_1 :

$$M_{21} = \frac{\lambda_{21}}{I_1} = \frac{N_2\Psi_{21}}{I_1}$$

- Now, the self-inductance of circuit-1 and circuit-2 is given by:

$$L_1 = \frac{\lambda_{11}}{I_1} = \frac{N_1\psi_1}{I_1}$$

$$L_2 = \frac{\lambda_{22}}{I_2} = \frac{N_2\psi_2}{I_2}$$

- The total energy in the magnetic field is the sum of the energies due to L_1 , L_2 and M_{12} (or M_{21}):

$$W_m = W_1 + W_2 + W_{12} = \frac{1}{2}L_1I_1^2 + \frac{1}{2}L_2I_2^2 \pm M_{12}I_1I_2$$

The +ve sign is taken if currents I_1 and I_2 flow such that the magnetic fields of the two circuits strengthen each other, otherwise the -ve sign is taken.

Inductors (contd.)

- We find the self-inductance L by taking the following steps.
 1. Choose a suitable coordinate system.
 2. Let the inductor carry current I .
 3. Determine \vec{B} from Biot-Savart Law (or from Ampere's circuital law if symmetry exists) and calculate ψ from $\psi = \int \vec{B} \cdot \overline{ds}$.
 4. Finally find L from $L = \frac{\lambda}{I} = \frac{N\psi}{I}$.

The mutual inductance between two circuits may be calculated by taking a similar procedure.

- In coaxial or a parallel-wire transmission line, the inductance produced by the flux internal to the conductor is called the internal inductance L_{in} while that produced by the flux external to it is called external inductance L_{ext} .
- The total inductance is given by:

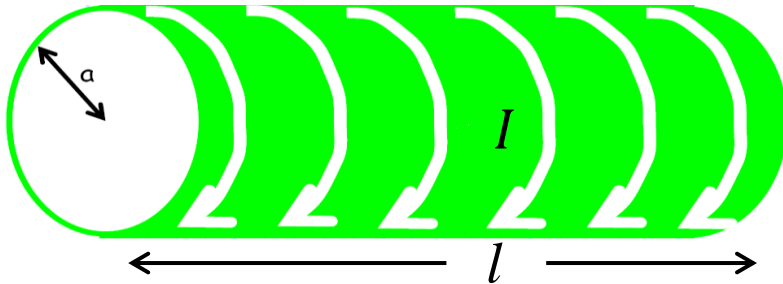
$$L = L_{in} + L_{ext}$$

Example – 1

- Calculate the self-inductance per unit length of an infinitely long solenoid.

Solution

- Consider a solenoid of length l and determine the magnetic field at a point along its axis.



Magnitude of magnetic field inside the solenoid along its axis per unit length:

$$B = \mu \frac{NI}{l}$$

- If S is the cross-section area of the solenoid then:

$$\psi = BS = \mu S \frac{NI}{l}$$

- Since this flux is only for a unit length of the solenoid, the linkage per unit length is:

$$\psi = BS = \mu S \frac{NI}{l}$$

- Therefore the inductance per unit length is:

$$L' = \frac{L}{l} = \frac{\lambda'}{I} = \frac{\mu SN^2}{l^2}$$

Magnetic Energy

- The potential energy in electrostatic field was derived as:

$$W_E = \frac{1}{2} \int \vec{D} \cdot \vec{E} dv = \frac{1}{2} \int \epsilon E^2 dv$$

One can derive similar expression for magnetostatic field

- Needless to say, the energy is stored in \vec{B} of the inductor \leftrightarrow therefore an appropriate expression should be either in terms of magnetic flux density \vec{B} or magnetic field \vec{H} .

- Expressions are:

$$w_m = \frac{1}{2} \mu H^2 = \frac{1}{2} \vec{B} \cdot \vec{H} = \frac{B^2}{2\mu}$$

- The total energy in a linear medium is:

$$W_m = \int w_m dv$$



$$W_m = \frac{1}{2} \int \vec{B} \cdot \vec{H} dv$$



$$W_m = \frac{1}{2} \int \mu H^2 dv$$

Example – 2

- Determine the self-inductance of a coaxial cable of inner radius a and outer radius b .

Solution

- The self-inductance could be determined using the 4-step process.
- Alternatively, use the energy concept.

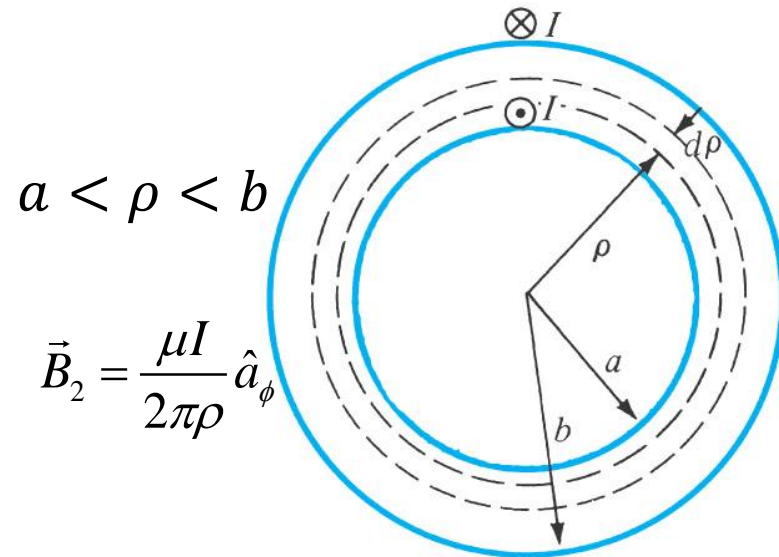
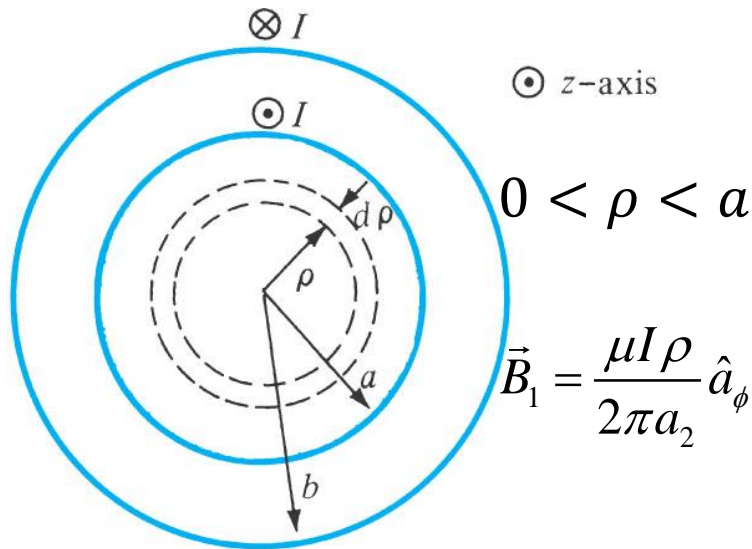
$$L = \frac{2W_m}{I^2}$$

where:

$$W_m = \frac{1}{2\mu} \int B^2 dv$$

Example – 2 (contd.)

- Let us consider the following configurations for co-axial line of length l .



- Therefore:

$$L_{in} = \frac{2}{I^2} \int \frac{B_1^2}{2\mu} dv = \frac{1}{I^2 \mu} \iiint_v \frac{\mu^2 I^2 \rho^2}{4\pi^2 a^4} \rho d\rho d\phi dz$$



$$L_{in} = \frac{\mu}{4\pi^2 a^4} \int_0^l dz \int_0^{2\pi} d\phi \int_0^a \rho^3 d\rho$$

$$\therefore L_{in} = \frac{\mu l}{8\pi}$$

Example – 2 (contd.)

- Then:

$$L_{ext} = \frac{2}{I^2} \int \frac{B_2^2}{2\mu} dv = \frac{1}{I^2 \mu} \iiint_v \frac{\mu^2 I^2}{4\pi^2 \rho^2} \rho d\rho d\phi dz$$



$$L_{ext} = \frac{\mu}{4\pi^2} \int_0^l dz \int_0^{2\pi} d\phi \int_a^b \frac{1}{\rho} d\rho$$



$$\therefore L_{ext} = \frac{\mu l}{2\pi} \ln \frac{b}{a}$$

- Therefore:

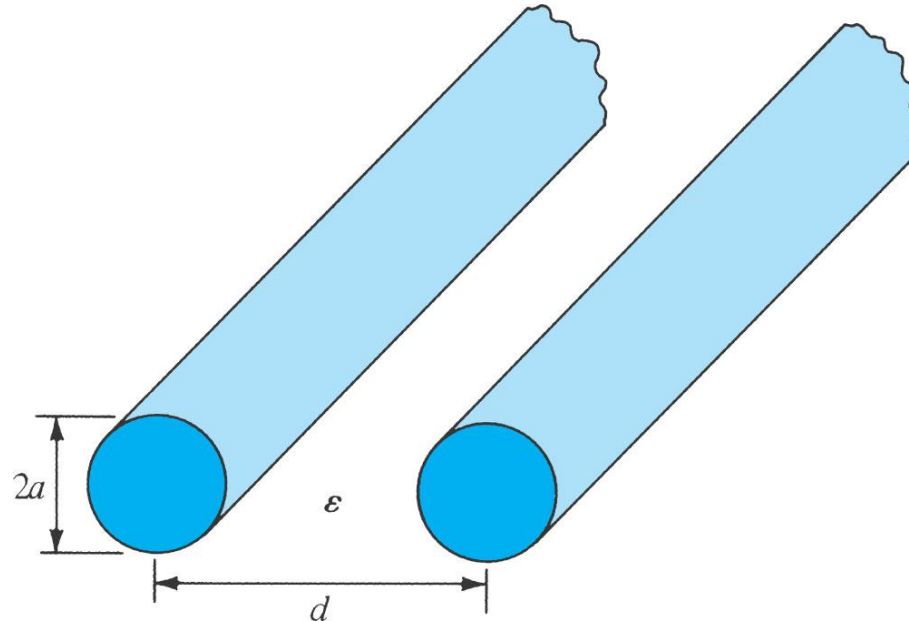
$$L = L_{in} + L_{ext}$$



$$L = \frac{\mu l}{2\pi} \left[\frac{1}{4} + \ln \frac{b}{a} \right]$$

Example – 3

- Determine the inductance per unit length of a two-wire transmission line with separation distance d shown below.



Example – 3 (contd.)

- The self-inductance could be determined using the 4-step process.
- Alternatively, use the energy concept.

Method – 1

For region $0 \leq \rho \leq a$:

$$\lambda_1 = \frac{\mu I l}{8\pi}$$

For region $a \leq \rho \leq d$:

$$\lambda_2 = \psi_2 = \int_{\rho=a}^{d-a} \int_{z=0}^l \frac{\mu I}{2\pi\rho} d\rho dz$$



$$\lambda_2 = \psi_2 = \frac{\mu I l}{2\pi} \ln \frac{d-a}{a}$$

Therefore the flux linkage produced by one wire is:

$$\lambda = \lambda_1 + \lambda_2 = \frac{\mu I l}{8\pi} + \frac{\mu I l}{2\pi} \ln \frac{d-a}{a}$$

By symmetry, the same amount of flux is produced by current $-I$ in second wire. Therefore:

$$L' = \frac{2L}{l} = \frac{\mu}{\pi} \left[\frac{1}{4} + \ln \frac{d-a}{a} \right]$$

Example – 3 (contd.)

Method – 2

$$\therefore L_{in} = \frac{\mu l}{8\pi}$$

- Now:

$$L_{ext} = \frac{2}{I^2} \int \frac{B^2}{2\mu} dv = \frac{1}{I^2 \mu} \iiint_v \frac{\mu^2 I^2}{4\pi^2 \rho^2} \rho d\rho d\phi dz$$



$$L_{ext} = \frac{\mu}{4\pi^2} \int_0^l dz \int_0^{2\pi} d\phi \int_a^{d-a} \frac{1}{\rho} d\rho$$



$$\therefore L_{ext} = \frac{\mu l}{2\pi} \ln \frac{d-a}{a}$$

- Since the two wires are symmetrical:

$$L = 2(L_{in} + L_{ext})$$



$$L = \frac{\mu l}{\pi} \left[\frac{1}{4} + \ln \frac{d-a}{a} \right]$$



$$\therefore L' = \frac{\mu}{\pi} \left[\frac{1}{4} + \ln \frac{d-a}{a} \right]$$

Magnetic Circuits

- In many engineering applications, we need to compute the magnetic fields for structures that lack sufficient symmetry for straight-forward application of Ampère's law. Then, we use an approximate method known as magnetic-circuit analysis.
- The advantage of the magnetic-circuit approach is that it can be applied to unsymmetrical magnetic cores with multiple coils.
- They are basically ferromagnetic structures (mostly Iron, Cobalt, Nickel alloys and compounds) with coils wound around them.
- Magnetic devices such as toroids, transformers, motors, generators, and relays may be considered as magnetic circuits.
- The concept of magnetic circuits is based on solving some magnetic field problems by using circuit approach.
- The analysis becomes simpler if analogy with electric circuit is exploited.

Magnetic Circuits (contd.)

Electric	Magnetic
Conductivity, σ	Permeability, μ
Field Intensity, \vec{E}	Field intensity, \vec{H}
Current, $I = \int \vec{j} \cdot \overline{ds}$	Magnetic Flux, $\psi = \int \vec{B} \cdot \overline{ds}$
Current Density, $\vec{j} = \sigma \vec{E} = \frac{I}{S}$	Flux Density, $\vec{B} = \mu \vec{H} = \frac{\psi}{S}$
Electromotive Force (emf), V	Magnetomotive Force (mmf), \mathcal{F}
Resistance, R	Reluctance, \mathcal{R}
Conductance	Permeance
Ohm's Law: $resistance = \frac{emf}{current}$	Ohm's Law: $reluctance = \frac{mmf}{magnetic\ flux}$
Kirchoff's Laws: $\sum I = 0$ $\sum V - \sum RI = 0$	Kirchoff's Laws: $\sum \psi = 0$ $\sum \mathcal{F} - \sum R\psi = 0$

Magnetic Circuit Definitions

- **Magnetomotive Force**
- The “driving force” that causes a magnetic field
- Symbol, \mathcal{F}
- Definition, $\mathcal{F} = \mathcal{N}I$
- Units, Ampere-turns, (A-t)
- The analogous quantity is Voltage (emf)

- **Flux Density:** The concentration of the lines of force in a magnetic circuit
- Symbol, B
- Definition, $B = \psi/A$
- Units, (Wb/m²), or T (Tesla)

- **Magnetic Field Intensity**
- mmf gradient, or mmf per unit length
- Symbol, H
- Definition, $\mathcal{H} = \mathcal{F}/\ell = \mathcal{N}I/\ell$
- Units, (A-t/m)

- **Reluctance:** The measure of “opposition” the magnetic circuit offers to the flux
- The analogous quantity of Resistance in an electrical circuit
- Symbol, \mathcal{R}
- Definition, $\mathcal{R} = \mathcal{F}/\psi$
- Units, (A-t/Wb)

Magnetic Circuit Definitions (contd.)

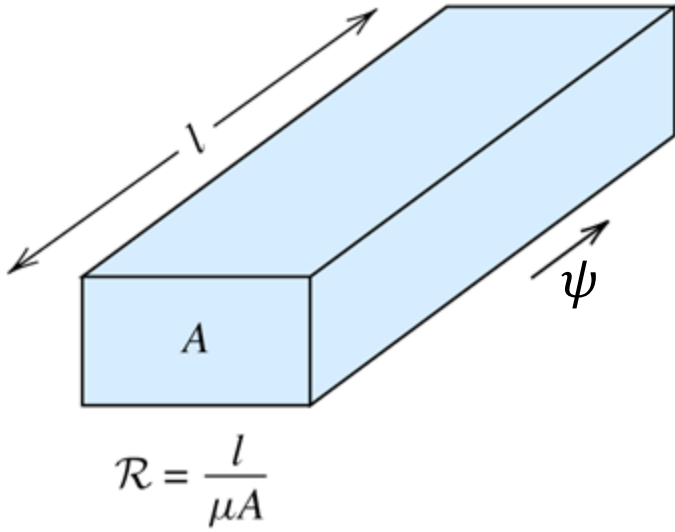
- **Permeability:** Relates flux density and field intensity
- Symbol, μ
- Definition, $\mu = B/H$
- Units, (Wb/A-t-m)

- **Permeability** of free space (air)
- Symbol, μ_0
- $\mu_0 = 4\pi \times 10^{-7}$ Wb/A-t-m

- **Relative Permeability:** Compares permeability of material with the permeability of free space (air)
- Symbol, μ_r
- $\mu_r = \mu/\mu_0$:Dimensionless

Magnetic Circuits (contd.)

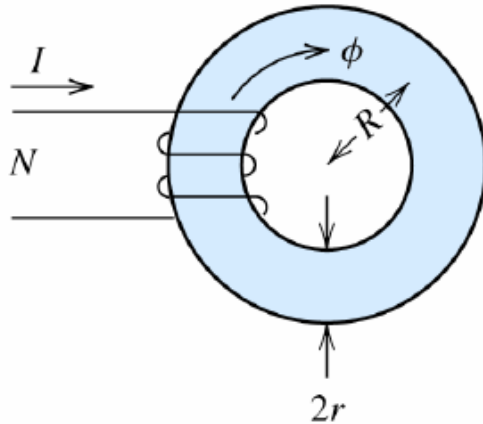
- reluctance of a path for magnetic flux



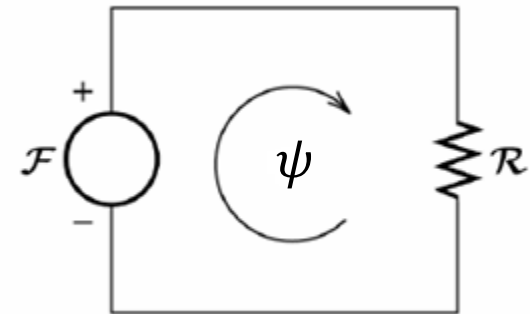
Reluctance \leftrightarrow Resistance

$$\mathcal{R} = \frac{1}{\mu} \left(\frac{l}{A} \right) \leftrightarrow R = \frac{1}{\sigma} \left(\frac{l}{A} \right)$$

Magnetic Circuits (contd.)



Corresponding
magnetic circuit



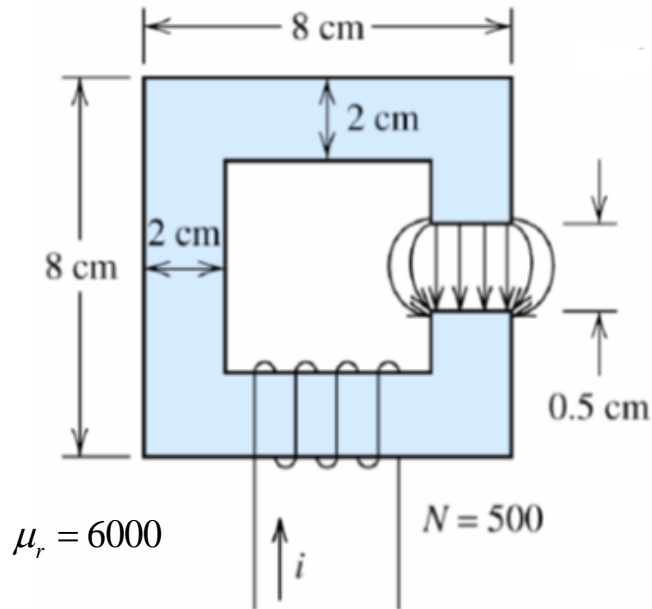
$$l = 2\pi R \quad A = \pi r^2$$

$$\mathcal{R} = \frac{1}{\mu} \left(\frac{l}{A} \right) = \frac{1}{\mu} \left(\frac{2\pi R}{\pi r^2} \right) = \frac{1}{\mu} \left(\frac{2R}{r^2} \right)$$

$$\mathcal{F} = NI$$

$$\psi = \frac{\mathcal{F}}{\mathcal{R}} = \frac{\mu N r^2 I}{2R}$$

Example – 4

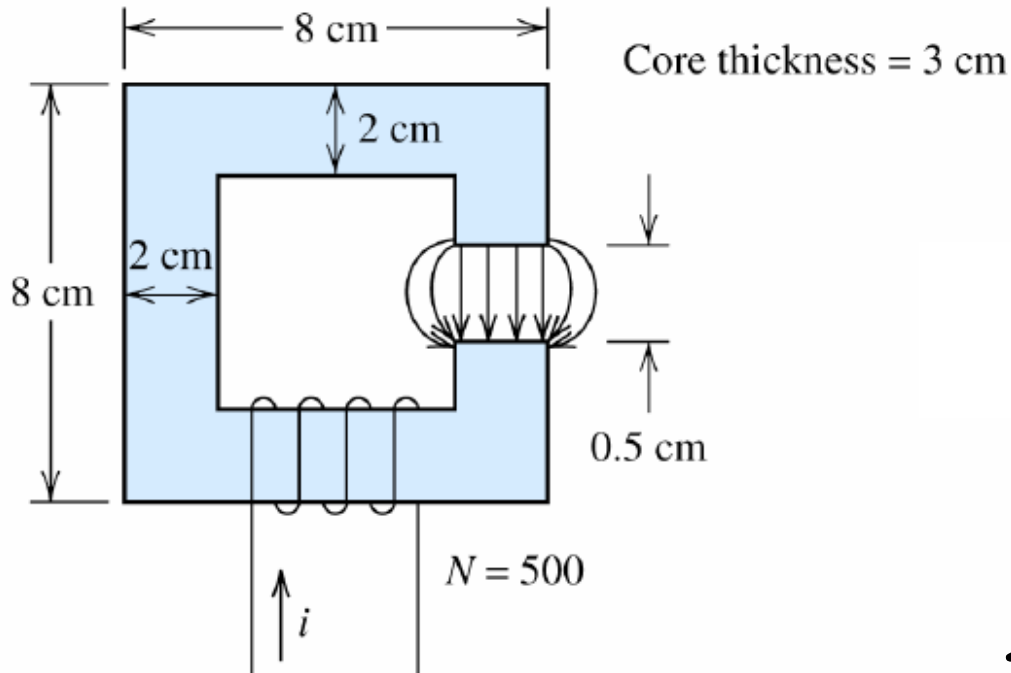


Core thickness: 3cm

- Find what current is required to generate a flux density of $B_{\text{gap}} = 0.25 \text{ T}$ in the air gap.

We approximately account for fringing by adding the length of the gap to the depth and width in computing effective gap area.

Example – 4 (contd.)



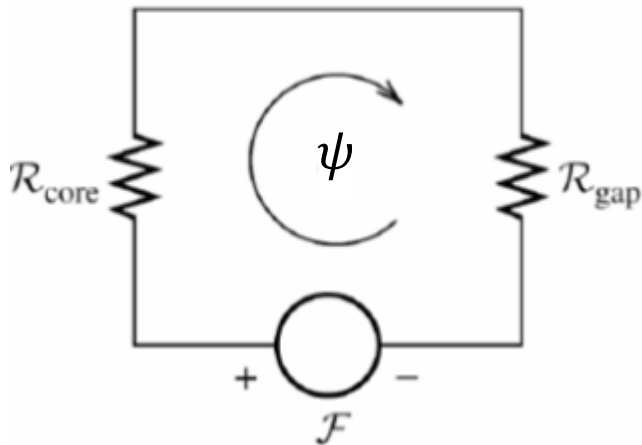
$$\begin{aligned} \mathcal{R}_{core} &= \frac{1}{\mu} \frac{l}{A} = \frac{1}{\mu_r \mu_0} \frac{(4 \times 6 - 0.5) \text{ cm}}{(2 \text{ cm})(3 \text{ cm})} \\ &= \frac{1}{(6000)(4\pi \times 10^{-7})} \frac{23.5 \times 10^{-2} \text{ m}}{6 \times 10^{-4} \text{ m}^2} \\ &= 5.195 \times 10^4 \end{aligned}$$

$$\begin{aligned} A_{gap} &= (2 \text{ cm} + 0.5 \text{ cm}) \times (3 \text{ cm} + 0.5 \text{ cm}) \\ &= 8.75 \times 10^{-4} \text{ m}^2 \end{aligned}$$

$$\mu_{gap} \approx \mu_0 = 4\pi \times 10^{-7}$$

$$\begin{aligned} \mathcal{R}_{gap} &= \frac{1}{4\pi \times 10^{-7}} \frac{0.5 \times 10^{-2} \text{ m}}{8.75 \times 10^{-4} \text{ m}^2} \\ &= 4.547 \times 10^6 \end{aligned}$$

Example – 4 (contd.)



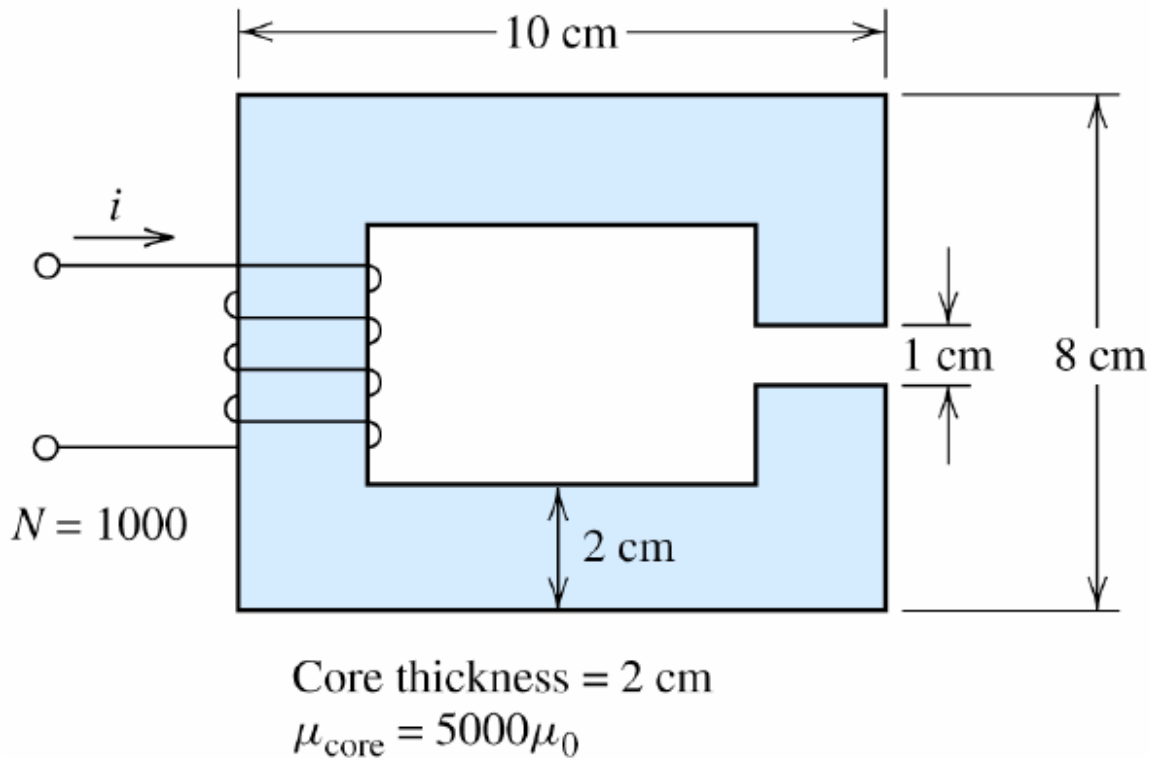
$$\begin{aligned}\mathcal{R}_{total} &= \mathcal{R}_{core} + \mathcal{R}_{gap} \\ &= 5.195 \times 10^4 + 4.547 \times 10^6 = 4.600 \times 10^6\end{aligned}$$

$$\begin{aligned}\psi &= B_{gap} A_{gap} = (0.25T)(8.75 \times 10^{-4} m^2) \\ &= 2.188 \times 10^{-4} Wb\end{aligned}$$

$$\begin{aligned}\mathcal{F} &= \psi \mathcal{R} = (2.188 \times 10^{-4})(4.600 \times 10^6) \\ &= 1006 A \text{ turns} \\ &= Ni\end{aligned}$$

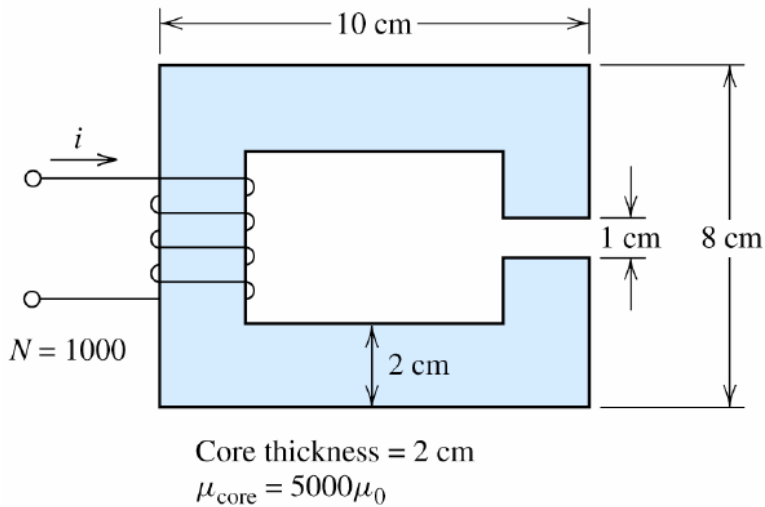
$$i = \frac{\mathcal{F}}{N} = \frac{1006 A \text{ turns}}{500 \text{ turns}} = 2.012 A$$

Example – 5



- Determine the current required to establish a flux density of 0.5T in the air gap

Example – 5 (contd.)



$$A_{\text{gap}} = (2\text{cm} + 1\text{cm}) \times (2\text{cm} + 1\text{cm})$$

$$= 9 \times 10^{-4} \text{ m}^2$$

$$\mu_{\text{gap}} \approx \mu_0 = 4\pi \times 10^{-7}$$

$$\mathcal{R}_{\text{gap}} = \frac{1}{4\pi \times 10^{-7}} \frac{1 \times 10^{-2} \text{ m}}{8.75 \times 10^{-4} \text{ m}^2}$$

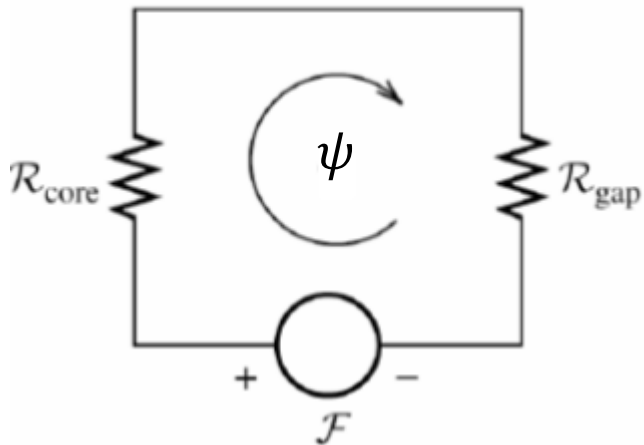
$$= 8.842 \times 10^6$$

$$\mathcal{R}_{\text{core}} = \frac{1}{\mu} \frac{l}{A} = \frac{1}{\mu_r \mu_0} \frac{(2 \times 8 + 2 \times 6 - 1) \text{ cm}}{(2 \text{ cm})(2 \text{ cm})}$$

$$= \frac{1}{(5000)(4\pi \times 10^{-7})} \frac{27 \times 10^{-2} \text{ m}}{4 \times 10^{-4} \text{ m}^2}$$

$$= 107.4 \times 10^3$$

Example – 5 (contd.)



$$\mathcal{R}_{total} = \mathcal{R}_{gap} + \mathcal{R}_{core}$$

$$= 8.842 \times 10^6 + 0.107 \times 10^6 \approx \mathcal{R}_{gap}$$

$$\psi = B_{gap} A_{gap} = (0.5T)(9 \times 10^{-4} m^2)$$

$$= 0.45 mWb$$

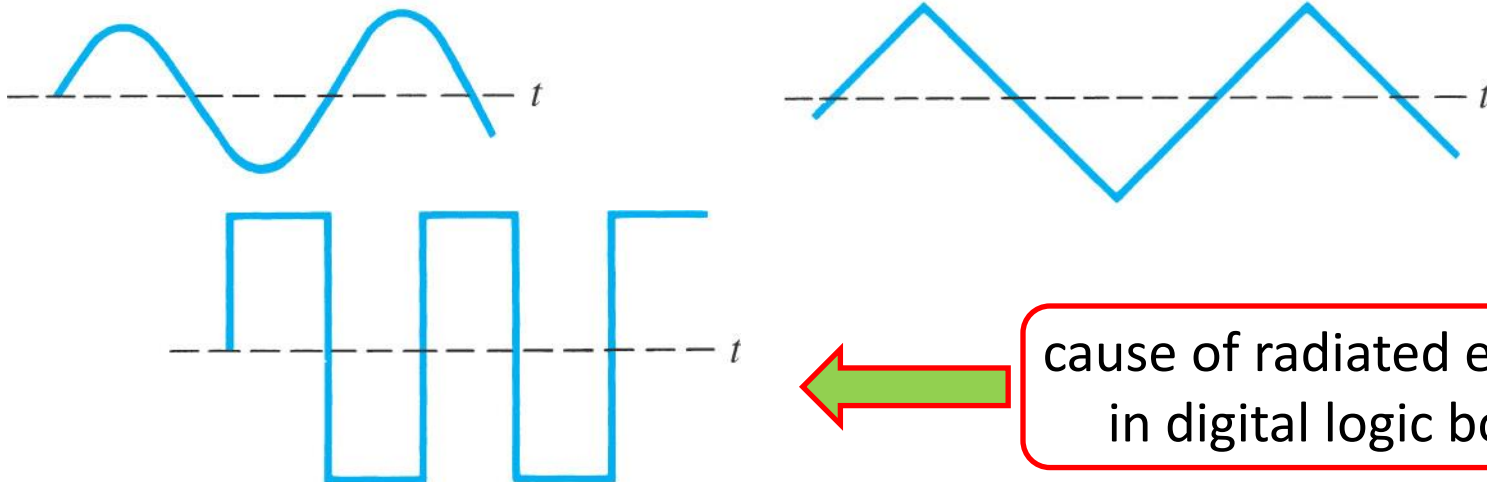
$$i = \frac{\mathcal{R}_{total} \psi}{N}$$

$$= \frac{(8.842 \times 10^6 + 0.107 \times 10^6)(0.45 \times 10^{-3})}{1000}$$

$$= 4.027 A$$

Electromagnetic Fields

- Stationary Charges \rightarrow Electrostatic Fields
- Steady Currents \rightarrow Magnetostatic Fields
- Time Varying Currents \rightarrow Electromagnetic Fields (or Waves)
- Any pulsating current will produce radiation (time-varying fields)



Faraday's Law

- Say instead of a static magnetic flux density, we consider a **time-varying** \vec{B} field (i.e., $\vec{B}(x, y, z, t)$).
- Recall that one of **Maxwell's** equations is:

$$\nabla \times \vec{E}(x, y, z) = -\frac{\partial \vec{B}(x, y, z, t)}{\partial t}$$

Yikes! The curl of the electric field is therefore **not zero** if the magnetic flux density is **time-varying**!

If the magnetic flux density is changing with time, the electric field will **not be conservative**!

Faraday's Law (contd.)

Q: What the heck does this equation mean ?!?

$$\nabla \times \vec{E}(x, y, z) = -\frac{\partial \vec{B}(x, y, z, t)}{\partial t}$$

A: Integrate both sides over some surface S:

$$\iint_S \nabla \times \vec{E}(x, y, z) \cdot \vec{ds} = -\frac{\partial}{\partial t} \iint_S \vec{B}(x, y, z, t) \cdot \vec{ds}$$

Stoke's
Theorem

$$\oint_L \vec{E}(x, y, z) \cdot \vec{dl} = -\frac{\partial}{\partial t} \iint_S \vec{B}(x, y, z, t) \cdot \vec{ds}$$

Note that $\oint \vec{E}(x, y, z) \cdot \vec{dl} \neq 0$

This equation is called **Faraday's Law of Induction.**

Q: Again, what does this **mean**?

A: It means that a time varying magnetic flux density $\vec{B}(x, y, z, t)$ can **induce** an electric field (and thus an electric potential difference)!

Faraday's Law describes the behavior of devices such as **generators, inductors, and transformers !**

Faraday's Law (contd.)

- Faraday discovered that an **induced potential difference** (or **electromotive force, emf**) in any closed circuit is equal to the time rate of change of the magnetic flux linkage by the circuit.

$$V_{emf} = -\frac{d\lambda}{dt} = -N \frac{d\Psi}{dt} = -N \frac{d}{dt} \int_s \vec{B} \cdot \vec{ds}$$

Derivative is a total time derivative and operates on the magnetic field \vec{B} as well as the differential surface area \vec{ds} .

- It is apparent that an *emf* can be generated in a closed loop under any of the three conditions
 - A time varying magnetic field linking a stationary loop; the induced *emf* is then called the *transformer emf*.
 - A moving loop with a time-varying surface area in a static field; the induced *emf* is then called *motional emf*.
 - A moving loop in a time-varying field \vec{B} .

Faraday's Law (contd.)

- The total *emf* is then given by:

$$V_{emf} = V_{emf}^{tr} + V_{emf}^m$$

- For stationary loop:

$$V_{emf}^m = 0$$

- For static \vec{B} :

$$V_{emf}^{tr} = 0$$

$$V_{emf} = -\frac{d\lambda}{dt} = -N \frac{d\Psi}{dt} = -N \frac{d}{dt} \int_s \vec{B} \cdot \vec{ds}$$

The negative sign in this expression shows that the induced voltage acts in such a way as to oppose the flux producing it.

This is known as ***Lenz's Law***.

It emphasizes that the direction of current flow in the circuit is such that the induced magnetic field produced by the induced current will oppose the change in the original magnetic field.

Stationary Loop in Time-Varying \vec{B}

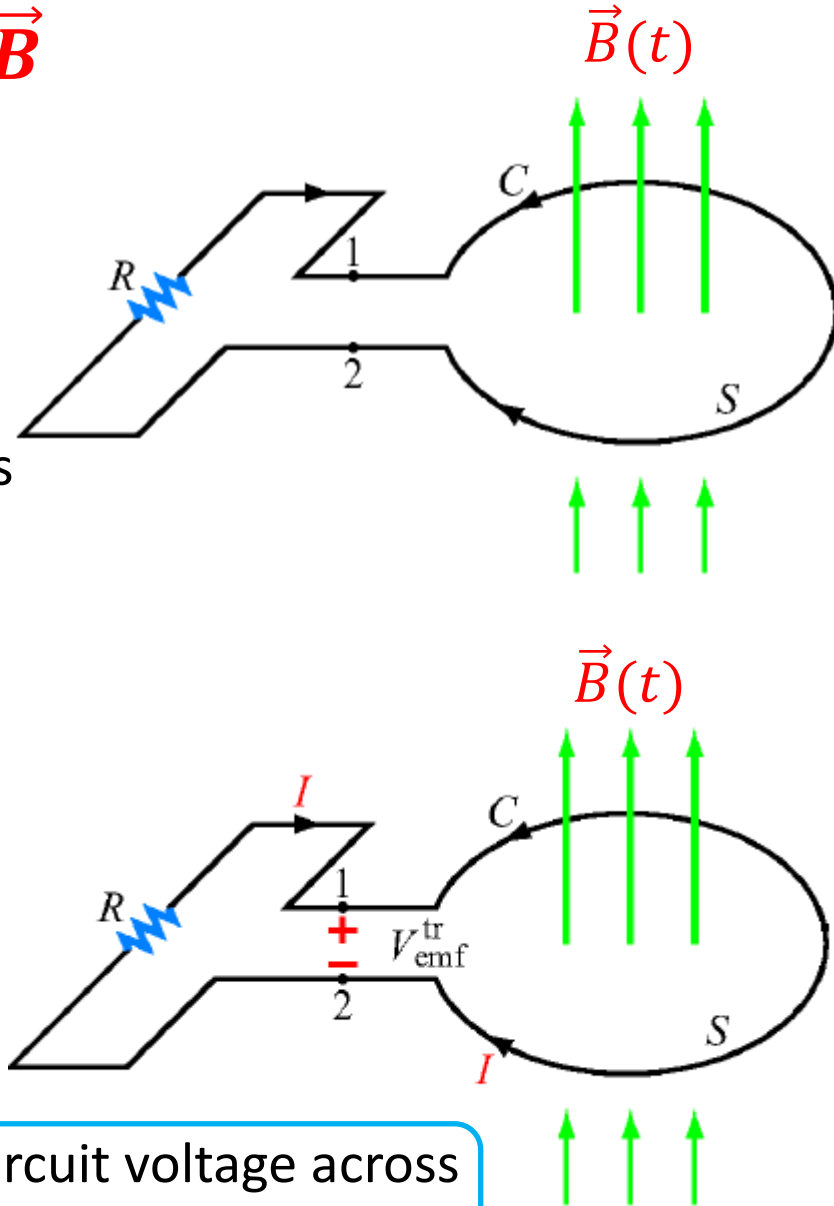
- Let us consider a stationary, single-turn, conducting, circular loop with contour C and surface area S placed in a time-varying magnetic field $\vec{B}(t)$.
- As stated, *emf* will be induced in this loop and its given by:

$$V_{emf}^{tr} = - \int_S \frac{\partial \vec{B}}{\partial t} \cdot \vec{ds}$$

- The *transformer emf* is the voltage difference that would appear across the small opening between terminals 1 and 2, even in the absence of the resistor R .

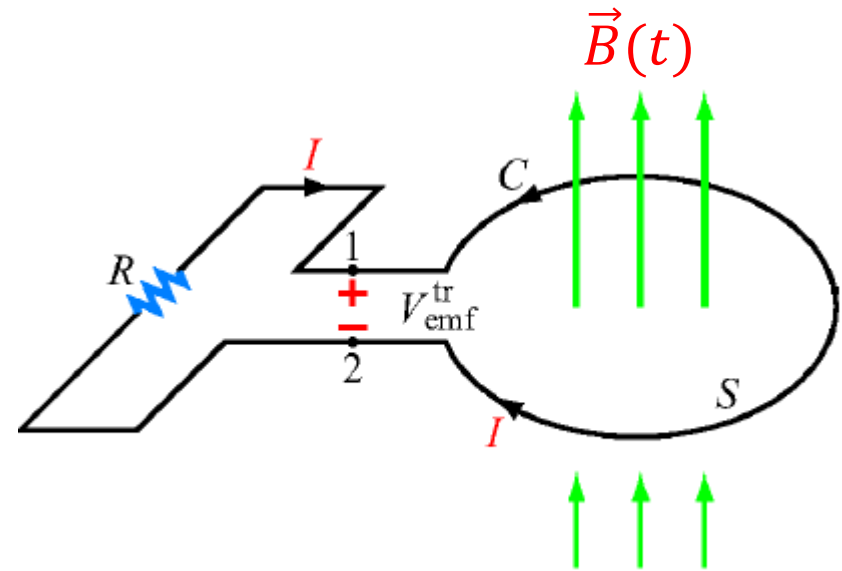
$$V_{emf}^{tr} = - \int_S \frac{\partial \vec{B}}{\partial t} \cdot \vec{ds}$$

V_{12} is the open-circuit voltage across the open ends of the loop



Stationary Loop in Time-Varying \vec{B} (contd.)

- The direction of \vec{ds} , the loops differential surface normal, can be chosen either upward or downward.
- These two choices are associated with the opposite designations of the polarities of terminals 1 and 2.
- The choice of direction of \vec{ds} and the polarity of emf is governed by right hand rule: If \vec{ds} points along the thumb of the right hand, then the directions of the contour C indicated by the four fingers is such that it always passes across the opening from the positive terminal to the negative terminal.

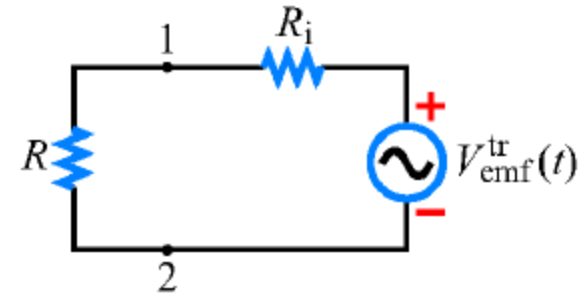


$$V_{emf}^{tr} = V_{12} = - \int_S \frac{\partial \vec{B}}{\partial t} \cdot \vec{ds}$$

Stationary Loop in Time-Varying \vec{B} (contd.)

- If the loop has an internal resistance R_i , the circuit can be represented equivalently as:
- Therefore the current I flowing through the circuit is:

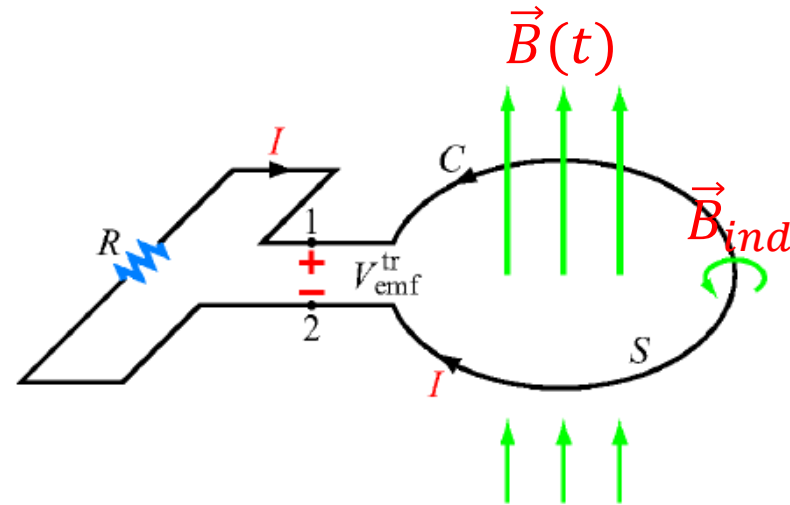
$$I = \frac{V_{emf}^{tr}}{R + R_i}$$



- The polarity of *emf* and hence the direction of I is governed by Lenz's law, which states that the current in the loop is always in a direction that opposes the change of magnetic flux $\psi(t)$ that produced I .

Stationary Loop in Time-Varying \vec{B} (contd.)

- The current I induces a magnetic field of its own, \vec{B}_{ind} , with a corresponding flux $\vec{\psi}_{ind}$.
- The direction of \vec{B}_{ind} is governed by right hand rule: If I is in a clockwise direction, then \vec{B}_{ind} points downward through S.
- Conversely**, if I is in counter clockwise direction, then \vec{B}_{ind} points upwards through S.
- If the original $\vec{B}(t)$ is increasing, means $\frac{d\psi}{dt} > 0$, then according to Lenz's law, I has to be in the direction shown in order for \vec{B}_{ind} to be in opposition to $\vec{B}(t)$.
- As a consequence, terminal 2 would be at higher potential and emf would have a negative value.



Stationary Loop in Time-Varying \vec{B} (contd.)

- However, if $\vec{B}(t)$ were to remain in the same direction but decrease in magnitude, means $\frac{d\psi}{dt} < 0$, then the current would have to reverse direction, and its induced field \vec{B}_{ind} would be in the same direction as $\vec{B}(t)$ so as to oppose the change (decrease) in $\vec{B}(t)$.

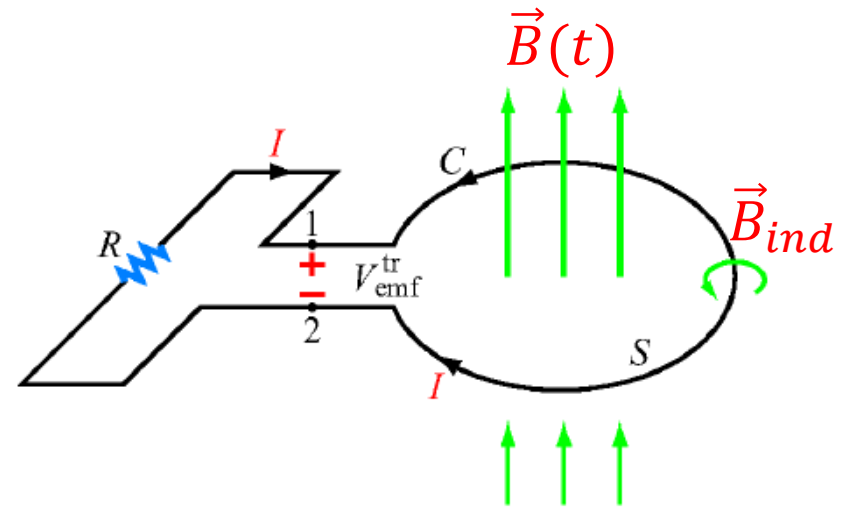
It is important to remember that \vec{B}_{ind} serves to oppose the change in $\vec{B}(t)$, and not necessarily $\vec{B}(t)$ itself.

Stationary Loop in Time-Varying \vec{B} (contd.)

Summary:

$$V_{emf}^{tr} = - \int_S \frac{\partial \vec{B}}{\partial t} \cdot \vec{ds}$$

$$V_{emf}^{tr} = \oint_C \vec{E} \cdot \vec{dl}$$



Its assumed that the
contour C is closed path
↔ Approximation

$$\oint_C \vec{E} \cdot \vec{dl} = - \int_S \frac{\partial \vec{B}}{\partial t} \cdot \vec{ds}$$

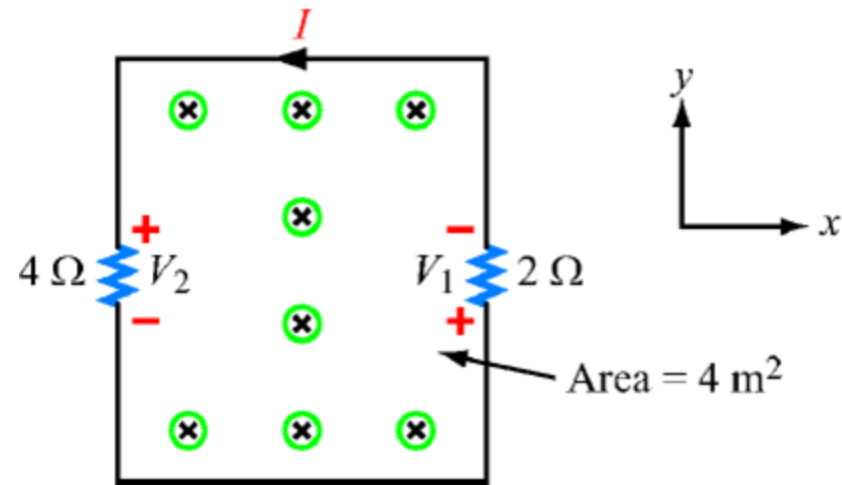
$$\int_S (\nabla \times \vec{E}) \cdot \vec{ds} = - \int_S \frac{\partial \vec{B}}{\partial t} \cdot \vec{ds}$$

$$\nabla \times \vec{E} = - \frac{\partial \vec{B}(t)}{\partial t}$$

The time varying magnetic field induces an electric field \vec{E} whose curl is equal to the negative of the time derivative of \vec{B} .

Example – 6

Determine voltages V_1 and V_2 across 2Ω and 4Ω resistors shown in the figure. The loop is located in xy – $plane$, its area is $4m^2$, the magnetic flux density is $\vec{B} = -\hat{a}_z 0.3t$ (T), and the internal resistance of the wire may be ignored.



- The flux flowing through the loop is:

$$\psi = \int_S \vec{B} \cdot d\vec{s} = \int_S (-0.3t\hat{a}_z) \cdot (ds\hat{a}_z)$$



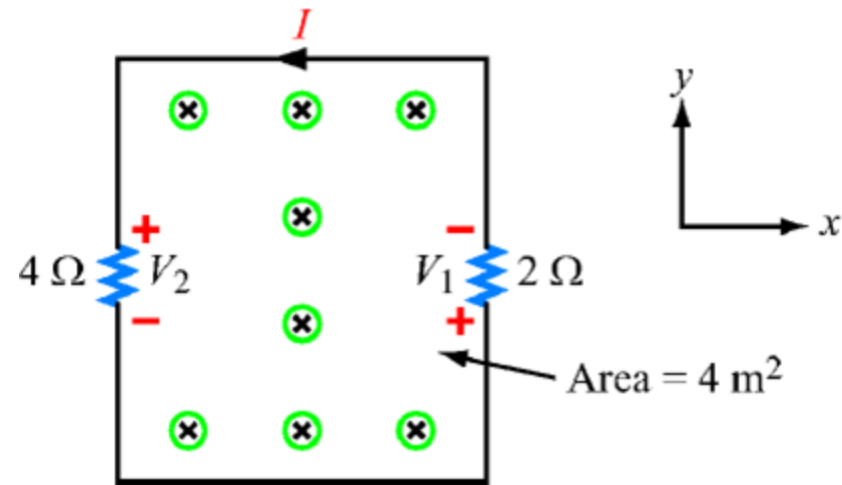
$$\therefore \psi = -0.3t \times 4 = -1.2t \text{ Wb}$$

- The corresponding transformer emf is: $V_{emf}^{tr} = -N \frac{d\Psi}{dt} = -(1) \frac{d\Psi}{dt} = -\frac{d\Psi}{dt} = 1.2V$

Given: the magnetic flux through the loop is along the $-z$ direction (into the page) and increases in magnitude with time t .

Example – 6 (contd.)

- According to Lenz's law, the induced current should be in a direction such that the magnetic flux density it induces (\vec{B}_{ind}) counteracts the direction of change of ψ .
- Therefore I must be in the direction as shown to get \vec{B}_{ind} in $+z$ direction.
- The total voltage of $1.2V$ is distributed across two resistors in series.



- As a consequence:
$$I = \frac{V_{emf}^{tr}}{R_1 + R_2} = \frac{1.2}{2 + 4} = 0.2\ \text{A}$$

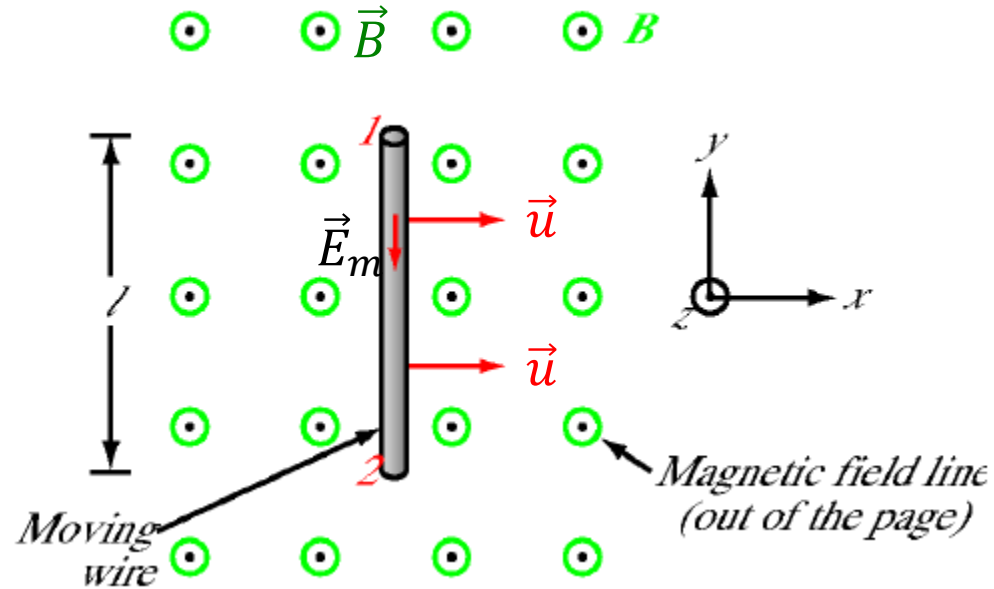
- Therefore:

$$V_1 = 0.2 \times 2 = 0.4\ \text{V}$$

$$V_2 = 0.2 \times 4 = 0.8\ \text{V}$$

Moving Conductor in a Static \vec{B}

- Let us consider a wire of length l moving across a static magnetic field $\vec{B} = \hat{a}_z B_0$ with constant velocity \vec{u} . The conducting wire contains free electrons.



- The magnetic force \vec{F}_m acting on a particle with charge q moving with velocity \vec{u} in a magnetic field \vec{B} is:

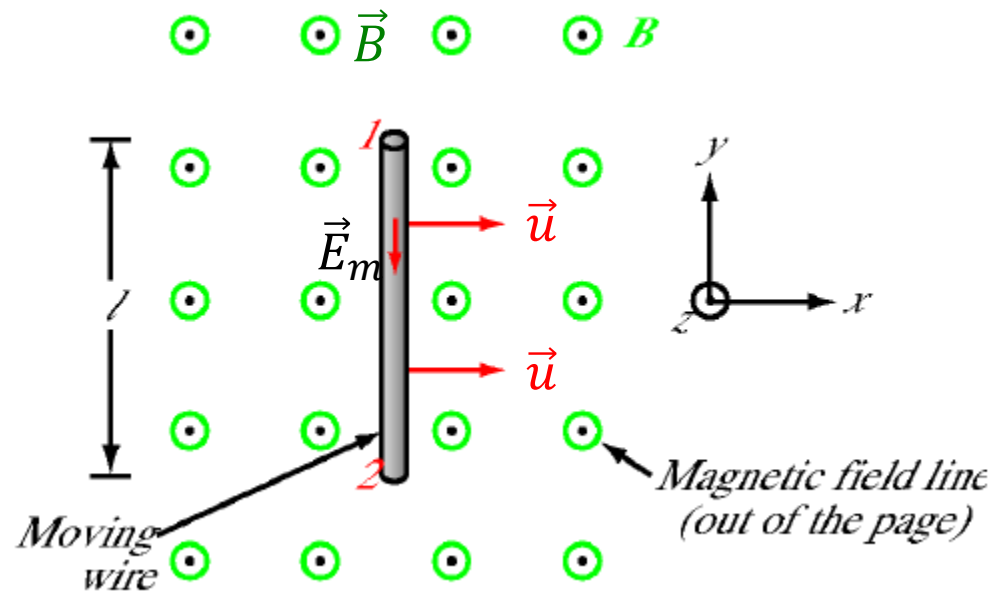
$$\vec{F}_m = q(\vec{u} \times \vec{B})$$

- This magnetic force is equivalent to the electrical force that would be exerted on the particle by the electric field \vec{E}_m given by:

$$\vec{E}_m = \frac{\vec{F}_m}{q} = \vec{u} \times \vec{B}$$

Moving Conductor in a Static \vec{B} (contd.)

- The field \vec{E}_m generated by the motion of the charged particle is called *motional electric field* and is orthogonal to both \vec{u} and \vec{B} .
- For our example, \vec{E}_m is along $-\hat{a}_y$.
- The magnetic force acting on the negatively charged electrons causes them to drift in the direction of $-\vec{E}_m$; i.e., toward the wire end label 1.
- The movement of electrons induces a voltage between ends 1 and 2.
- The induced voltage is called *motional emf*.



- motional emf* is defined as:

$$V_{emf}^m = V_{12} = \int_2^1 \vec{E}_m \cdot d\vec{l} = \int_2^1 (\vec{u} \times \vec{B}) \cdot d\vec{l}$$

Moving Conductor in a Static \vec{B} (contd.)

- For the conducting wire:

$$\vec{u} \times \vec{B} = u \hat{a}_x \times \hat{a}_z B_0 = -\hat{a}_y u B_0$$

$$\vec{dl} = \hat{a}_y dl$$

- Therefore: $V_{emf}^m = V_{12} = -u B_0 l$

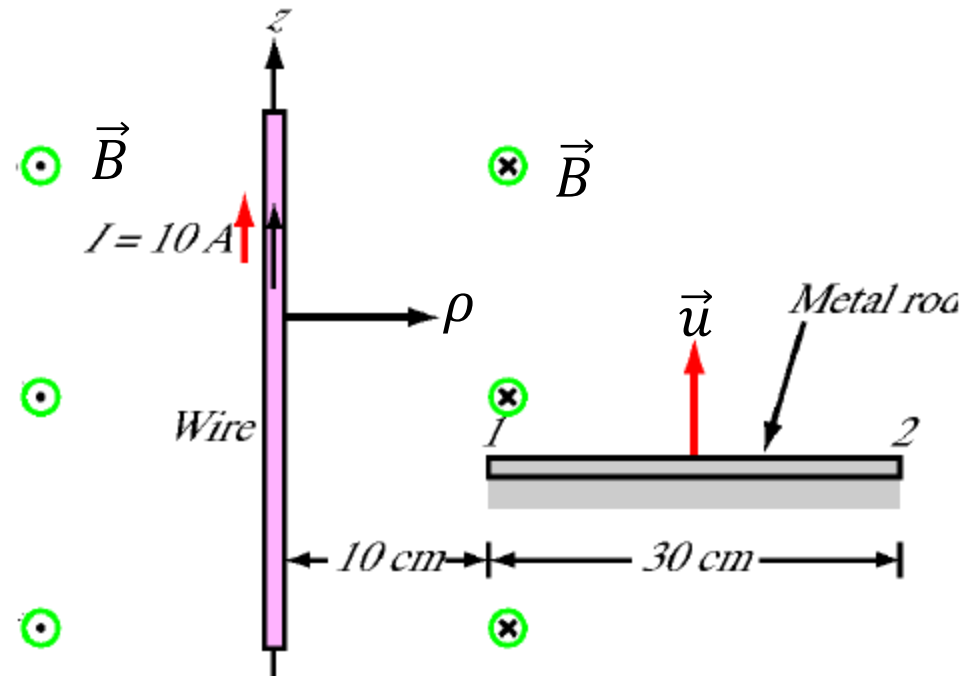
- In general, if any segment of a closed circuit with contour C moves with a velocity \vec{u} across a static magnetic field \vec{B} , then the induced *motional emf* is:

$$V_{emf}^m = \oint_C (\vec{u} \times \vec{B}) \cdot \vec{dl}$$

Only those segments of the circuit that cross magnetic field lines contribute to *motional emf*.

Example – 7

- The wire shown in the figure carries a current $I = 10A$. A 30-cm long metal rod moves with a constant velocity $\vec{u} = 5\hat{a}_z$ m/s. Find V_{12} .



- The current I induces a magnetic field:

$$B = \frac{\mu_0 I}{2\pi\rho} \hat{a}_\phi$$

Where, ρ is the radial distance from the wire and \hat{a}_ϕ is into the page at the rod side.

Example – 7 (contd.)

- The movement of the rod in the presence of the field \vec{B} induces a *motional emf* given by:

$$V_{emf}^m = V_{12} = \int_{40cm}^{10cm} (\vec{u} \times \vec{B}) \cdot \overline{d\vec{l}}$$

$$V_{12} = \int_{40cm}^{10cm} \left(5\hat{a}_z \times \frac{\mu_0 I}{2\pi\rho} \hat{a}_\phi \right) \cdot (\hat{a}_\rho d\rho)$$

$$\therefore V_{12} = 13.9(\mu V)$$