

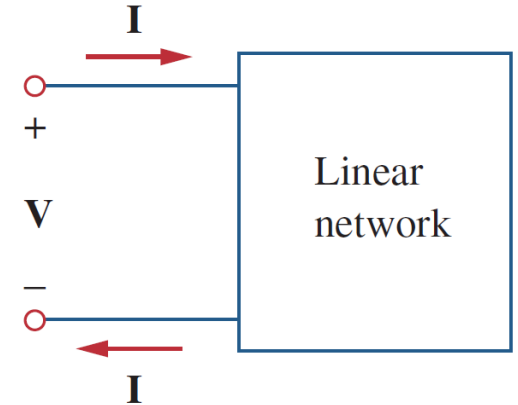
Lecture – 23

Date: 30.10.2017

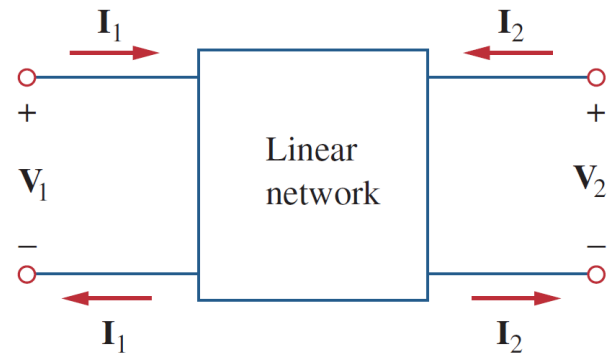
- Multi-port networks
- Impedance and Admittance Matrix
- Lossless and Reciprocal Networks

Introduction

- A pair of terminals through which a current may enter or leave a network is known as a *port*.
- Two-terminal devices or elements (such as resistors, capacitors, and inductors) result in one-port networks. Most of the circuits we have dealt with so far are two-terminal or one-port circuits.



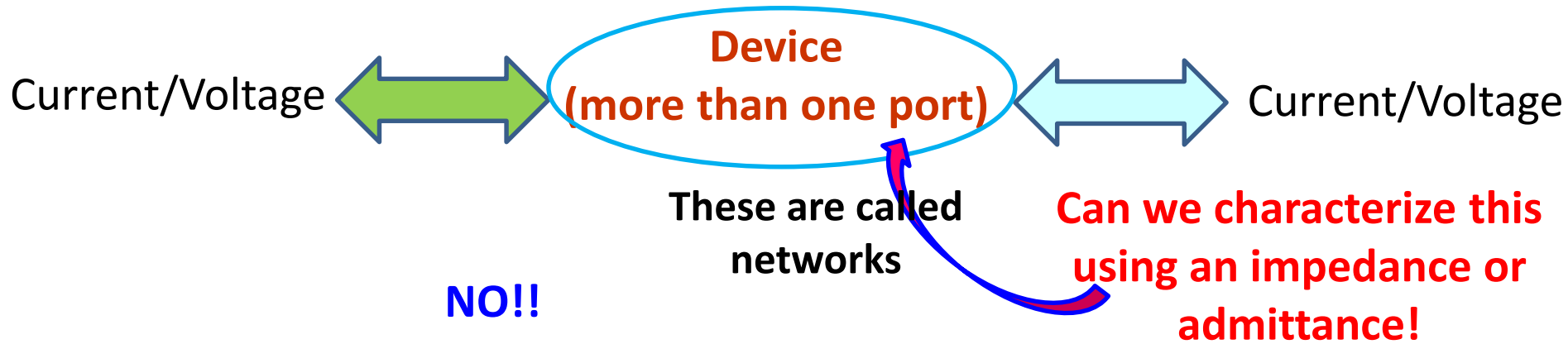
- So far, considered the voltage across or current through a single pair of terminals—such as the two terminals of a resistor, capacitor, inductor.
- We have also studied four-terminal or two-port circuits involving op amps, and transformers.



A **two-port network** is an electrical network with two separate ports for input and output.

2-port Networks

- Requirement of Matrix Formulation



What is the way?

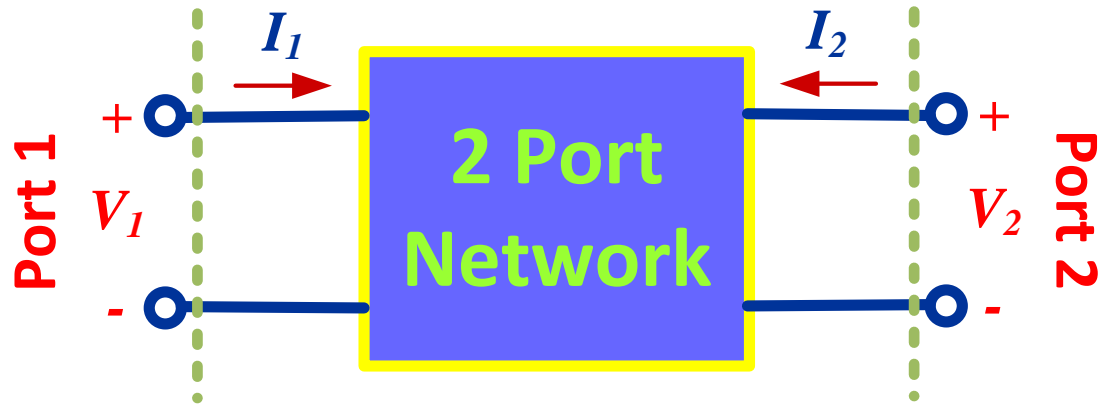
Impedance or Admittance Matrix. Right?

In principle, N by N impedance matrix completely characterizes a linear N-port device. Effectively, the impedance matrix defines a multi-port device the way a Z_L describes a single port device (e.g., a resistor)

Linear networks can be completely characterized by parameters measured at the network ports without knowing the content of the networks.

Multiport Networks

- Networks can have any number of ports – however, analysis of a 2-port, 3-port or 4-port network is sufficient to explain the theory and the associated concepts.



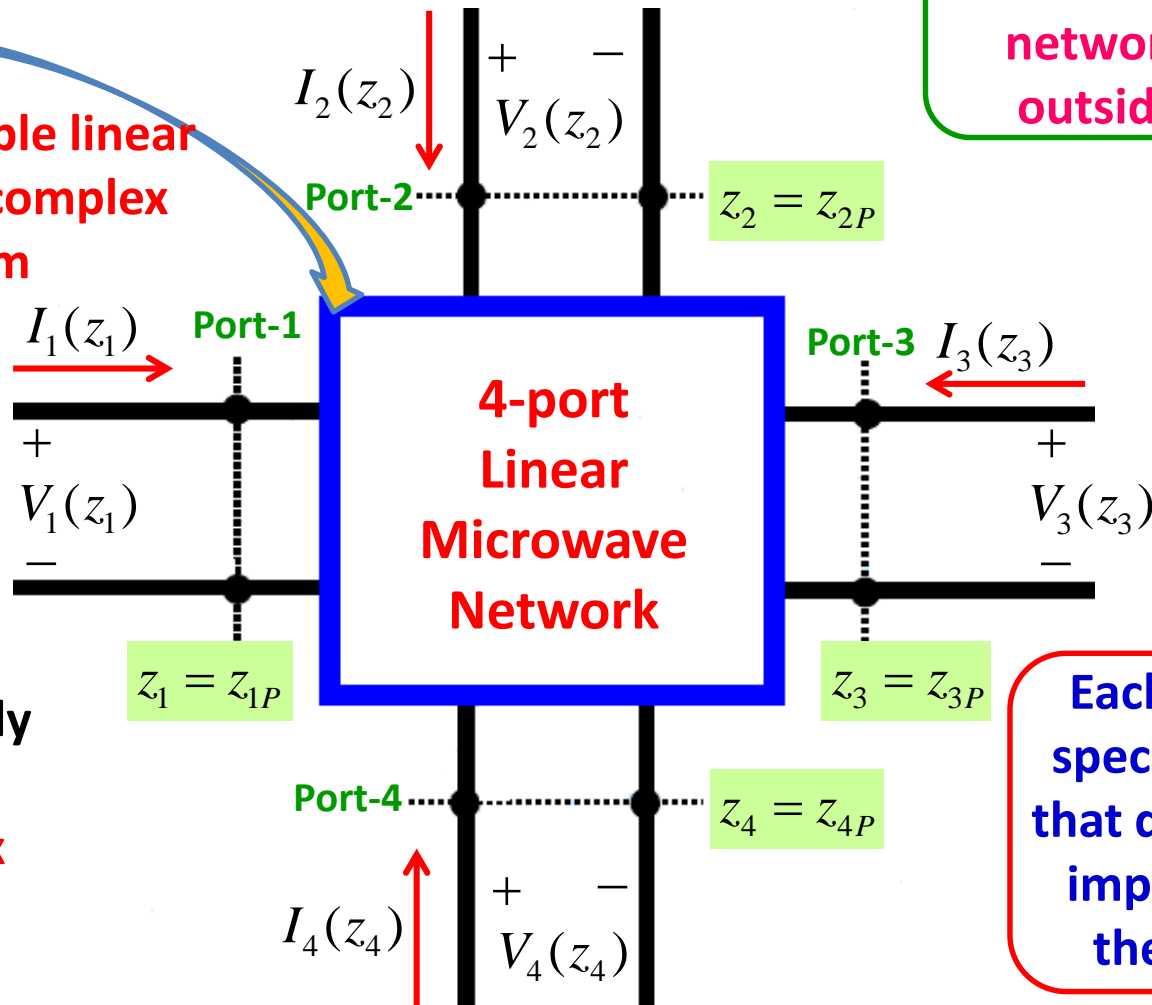
- The ports can be characterized with many parameters (Z , Y , h , S , $ABCD$). Each has a specific advantage.
- For 2-port Network, each parameter set is related to 4 variables:
 - 2 independent variables for excitation
 - 2 dependent variables for response

The Impedance Matrix

- Let us consider the following 4-port network:

Four identical cables used to connect this network to the outside world

This could be a simple linear device or a large/complex linear system



Either way, the network can be fully described by its impedance matrix

Each cable has specific location that defines input impedances to the network

The arbitrary locations are known as ports of the network

The Impedance Matrix (contd.)

- In principle, the current and voltages at the port- n of networks are given as:

$$V_n(z_n = z_{nP}) \quad I_n(z_n = z_{nP})$$

- the simplified formulations are:

$$V_n = V_n(z_n = z_{nP}) \quad I_n = I_n(z_n = z_{nP})$$

- If we want to say that there exists a non-zero current at port-1 and zero current at all other ports then we can write as:

$$I_1 \neq 0 \quad I_2 = I_3 = I_4 = 0$$

- In order to define the elements of impedance matrix, there will be need to measure/determine the associated voltages and currents at the respective ports. Suppose, if we measure/determine current at port-1 and then voltage at port-2 then we can define:

$$Z_{21} = \frac{V_2}{I_1}$$

Trans-impedance

- Similarly, the trans-impedance parameters Z_{31} and Z_{41} are:

$$Z_{31} = \frac{V_3}{I_1} \quad Z_{41} = \frac{V_4}{I_1}$$

- We can define other trans-impedance parameters such as Z_{34} as the ratio between the complex values I_4 (into port-4) and V_3 (at port-3), given that the currents at all other ports (1, 2, and 3) are zero.

The Impedance Matrix (contd.)

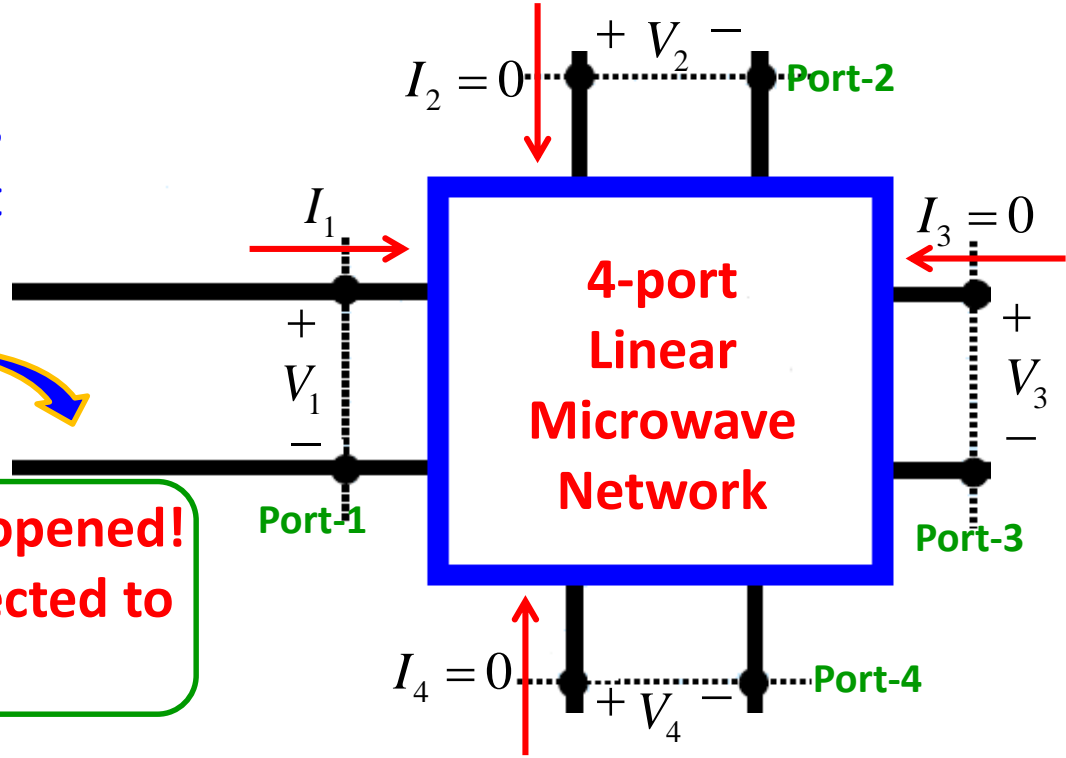
- Therefore, the more generic form of trans-impedance is:

$$Z_{mn} = \frac{V_m}{I_n} \quad (\text{given that } I_k = 0 \text{ for all } k \neq n)$$

How do we ensure that all but one port current is zero?

- Open the ports where the current needs to be zero

The ports should be opened!
not the cables connected to the ports



- We can then define the respective trans-impedances as:

$$Z_{mn} = \frac{V_m}{I_n} \quad (\text{given that all ports } k \neq n \text{ are open})$$

The Impedance Matrix (contd.)

- Once we have defined the trans-impedance terms by opening various ports, it is time to formulate the impedance matrix
- Since the network is **linear**, the **voltage at any port** due to **all the port currents** is simply the coherent **sum** of the voltage at that port due to **each** of the currents

- For example, the voltage at **port-3** is: $V_3 = Z_{34}I_4 + Z_{33}I_3 + Z_{32}I_2 + Z_{31}I_1$

- Therefore we can generalize the voltage for **N-port** network as:
$$V_m = \sum_{n=1}^N Z_{mn} I_n \quad \Rightarrow \quad \mathbf{V} = \mathbf{Z}\mathbf{I}$$

- Where **I** and **V** are vectors given as:
$$\mathbf{V} = [V_1, V_2, V_3, \dots, V_N]^T \quad \mathbf{I} = [I_1, I_2, I_3, \dots, I_N]^T$$

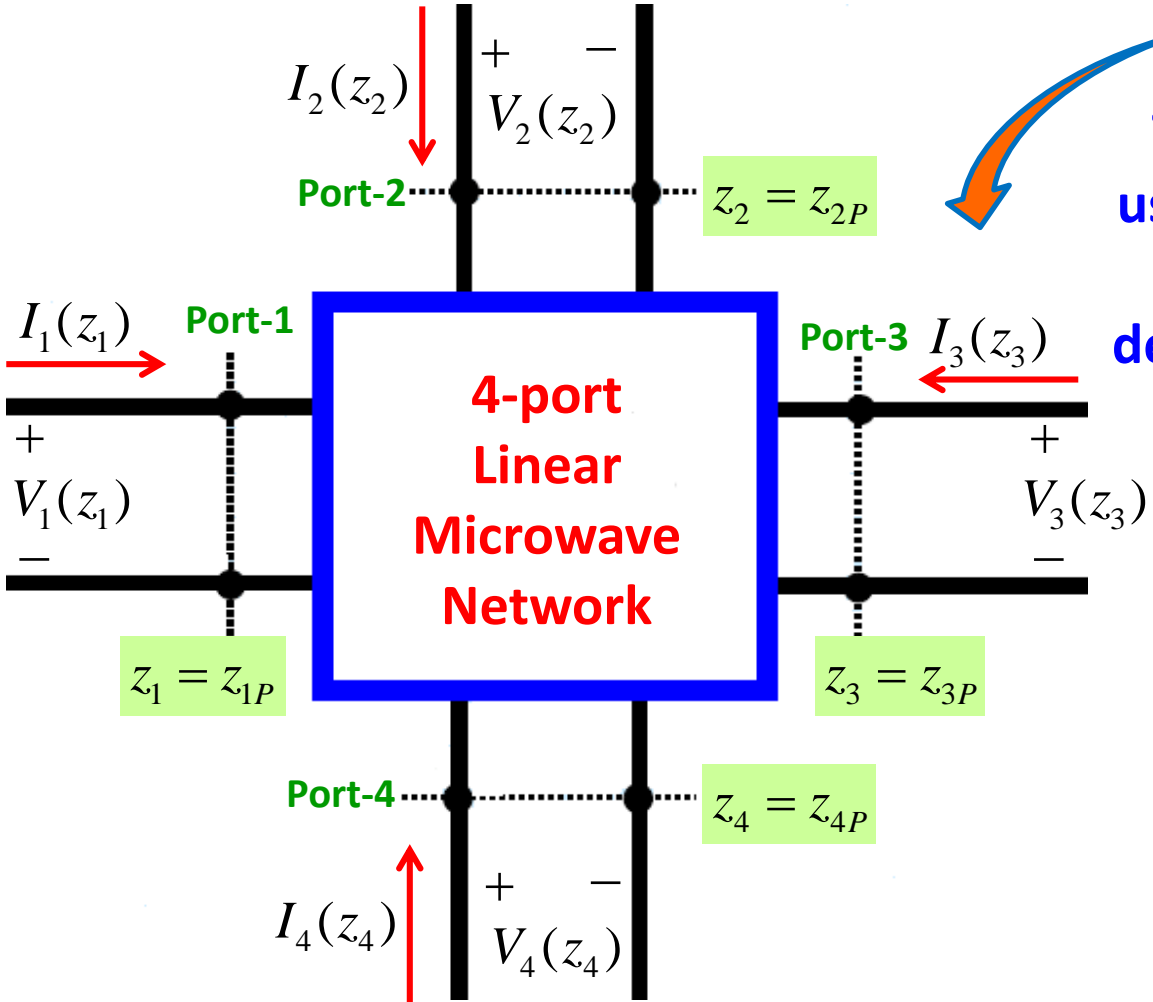
- The term **Z** is matrix given by:

$$\mathbf{Z} = \begin{bmatrix} Z_{11} & Z_{12} & \dots & Z_{1n} \\ Z_{21} & & & \vdots \\ \vdots & & & \\ Z_{m1} & Z_{m2} & \dots & Z_{mn} \end{bmatrix}$$

Impedance Matrix

The Admittance Matrix

- Let us consider the 4-port network again:



This can be characterized using admittance matrix – if currents are taken as dependent variables instead of voltages

The elements of admittance matrix are called trans-admittance parameters Y_{mn}

The Admittance Matrix (contd.)

- The trans-admittances Y_{mn} are defined as:

$$Y_{mn} = \frac{I_m}{V_n}$$

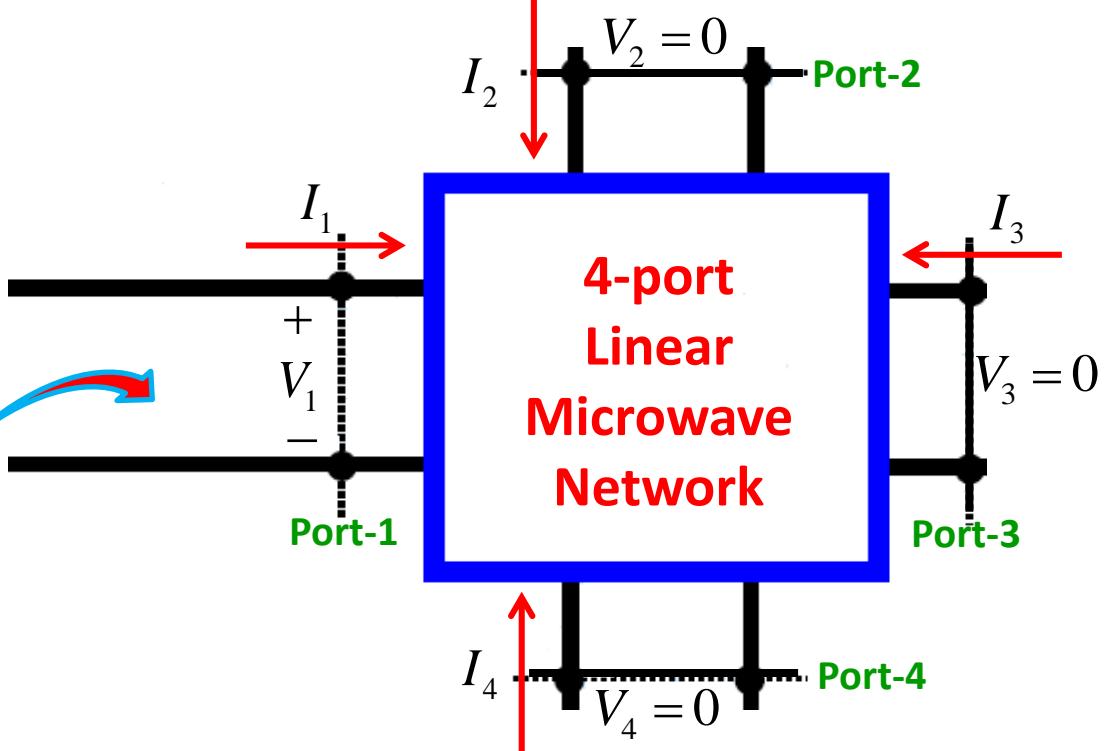
(given that $V_k = 0$ for all $k \neq n$)

Important

$$Y_{mn} \neq \frac{1}{Z_{mn}}$$

- It is apparent that the voltage at all but one port must be equal to zero. This can be ensured by short-circuiting the voltage ports.

The ports should be short-circuited! not the cable connected to the ports



- Now, since the network is **linear**, the **current at any one port** due to **all the port voltages** is simply the coherent **sum** of the currents at that port due to **each** of the port voltages.

The Admittance Matrix (contd.)

- For example, the current at **port-3** is:
- Therefore we can generalize the current for **N-port** network as:
- Where **I** and **V** are vectors given as:

$$I_3 = Y_{34}V_4 + Y_{33}V_3 + Y_{32}V_2 + Y_{31}V_1$$

$$I_m = \sum_{n=1}^N Y_{mn} V_n$$



$$\Rightarrow \mathbf{I} = \mathbf{YV}$$

$$\mathbf{V} = [V_1, V_2, V_3, \dots, V_N]^T$$

$$\mathbf{I} = [I_1, I_2, I_3, \dots, I_N]^T$$

- The term **Y** is matrix given by:

$$\mathbf{Y} = \begin{bmatrix} Y_{11} & Y_{12} & \dots & Y_{1n} \\ Y_{21} & & & \vdots \\ \vdots & & & \\ Y_{m1} & Y_{m2} & \dots & Y_{mn} \end{bmatrix}$$



Admittance Matrix

- The values of elements in the admittance matrix are frequency dependents and often it is advisable to describe admittance matrix as:

$$\mathbf{Y}(\omega) = \begin{bmatrix} Y_{11}(\omega) & Y_{12}(\omega) & \dots & Y_{1n}(\omega) \\ Y_{21}(\omega) & & & \vdots \\ \vdots & & & \\ Y_{m1}(\omega) & Y_{m2}(\omega) & \dots & Y_{mn}(\omega) \end{bmatrix}$$

The Admittance Matrix (contd.)



You said that:

$$Y_{mn} \neq \frac{1}{Z_{mn}}$$

Is there any relationship between admittance and impedance matrix of a given device?

Answer: Let us see if we can figure it out!

- Recall that we can determine the inverse of a matrix. Denoting the matrix inverse of the admittance matrix as Y^{-1} , we find:

$$\mathbf{I} = \mathbf{YV}$$

$$\Rightarrow \mathbf{Y}^{-1}\mathbf{I} = \mathbf{Y}^{-1}(\mathbf{YV}) \quad \longrightarrow \quad \mathbf{Y}^{-1}\mathbf{I} = (\mathbf{Y}^{-1}\mathbf{Y})\mathbf{V} \quad \longrightarrow \quad \mathbf{Y}^{-1}\mathbf{I} = \mathbf{V}$$

- We also know:

$$\mathbf{V} = \mathbf{ZI}$$

$$\mathbf{Z} = \mathbf{Y}^{-1} \quad \text{OR} \quad \mathbf{Y} = \mathbf{Z}^{-1}$$

Reciprocal and Lossless Networks

- We can **classify** multi-port devices or networks as either **lossless** or **lossy**; **reciprocal** or **non-reciprocal**. Let's look at each classification individually.

Lossless Network

- A **lossless** network or device is simply one that **cannot** absorb power. This does **not** mean that the delivered power at **every port** is zero; rather, it means the total power flowing **into** the **device** must equal the total power **exiting** the **device**.
- A lossless device exhibits an impedance matrix with an interesting **property**. Perhaps not surprisingly, we find for a lossless device that the **elements** of its impedance matrix will be **purely reactive**:

$$\text{Re}(Z_{mn}) = 0$$

For a lossless device
- If the device is lossy, then the elements of the impedance matrix must have **at least** one element with a real (i.e., resistive) component.
- Furthermore, we can similarly say that if the elements of an **admittance** matrix are **all** purely imaginary (i.e., $\text{Re}\{Y_{mn}\} = 0$), then the device is lossless.

Reciprocal and Lossless Networks (contd.)

Reciprocal Network

- Ideally, most **passive, linear** components will turn out to be **reciprocal**— regardless of whether the designer **intended** it to be or not!
- Reciprocity is a tremendously important characteristic, as it greatly **simplifies** an impedance or admittance matrix!
- Specifically, we find that a reciprocal device will result in a **symmetric** impedance and admittance **matrix**, meaning that:

$$Z_{mn} = Z_{nm}$$

$$Y_{mn} = Y_{nm}$$

For a reciprocal device

- **For example**, we find for a reciprocal device that $Z_{23} = Z_{32}$, and $Y_{12} = Y_{21}$.

Reciprocal and Lossless Networks (contd.)

$$\mathbf{Z} = \begin{bmatrix} j2 & 0.1 & j3 \\ -j & -1 & 1 \\ 4 & -2 & 0.5 \end{bmatrix}$$

neither
lossless nor
reciprocal

lossless,
but not
reciprocal

$$\mathbf{Z} = \begin{bmatrix} j2 & j0.1 & j3 \\ -j & -j1 & j1 \\ j4 & -j2 & j0.5 \end{bmatrix}$$

$$\mathbf{Z} = \begin{bmatrix} j2 & -j & 4 \\ -j & -1 & -j2 \\ 4 & -j2 & j0.5 \end{bmatrix}$$

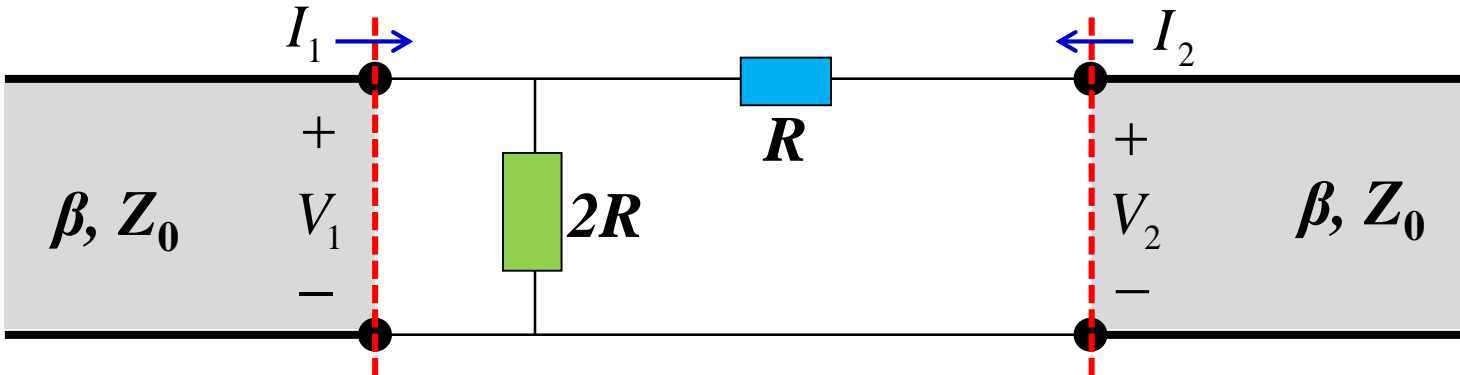
reciprocal,
but not
lossless

lossless
and
reciprocal

$$\mathbf{Z} = \begin{bmatrix} j2 & -j & j4 \\ -j & -j & -j2 \\ j4 & -j2 & j0.5 \end{bmatrix}$$

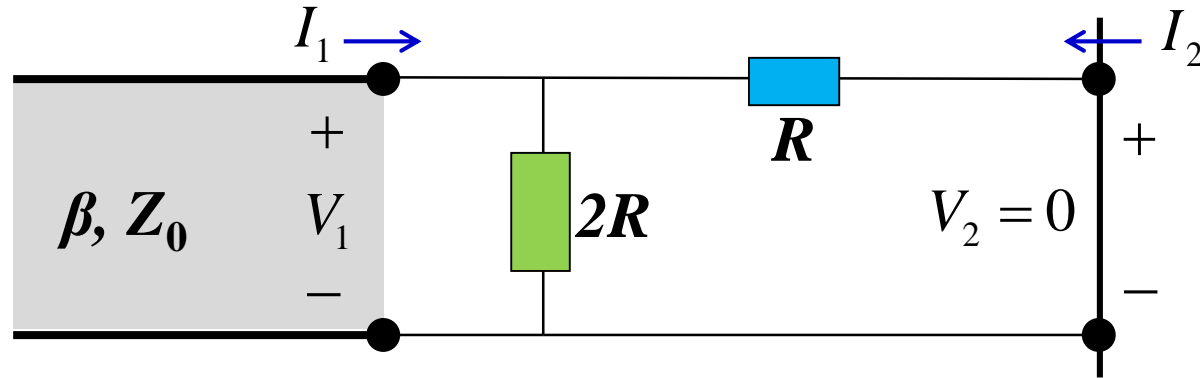
Example – 1

- determine the **Y matrix** of this two-port device.



Example – 1 (contd.)

Step-1: Place a **short** at port 2



Step-2: Determine currents I_1 and I_2

- Note that **after** the short was placed at port 2, both resistors are in **parallel**, with a potential V_1 across each

Therefore current I_1 is



$$I_1 = \frac{V_1}{2R} + \frac{V_1}{R} = \frac{3V_1}{2R}$$

- The current I_2 equals the portion of current I_1 through R but with opposite sign

$$I_2 = -\frac{V_1}{R}$$

Example – 1 (contd.)

Step-3: Determine the trans-admittances Y_{11} and Y_{21}

$$Y_{11} = \frac{I_1}{V_1} = \frac{3}{2R}$$

$$Y_{21} = \frac{I_2}{V_1} = -\frac{1}{R}$$

Note that Y_{21} is real and negative

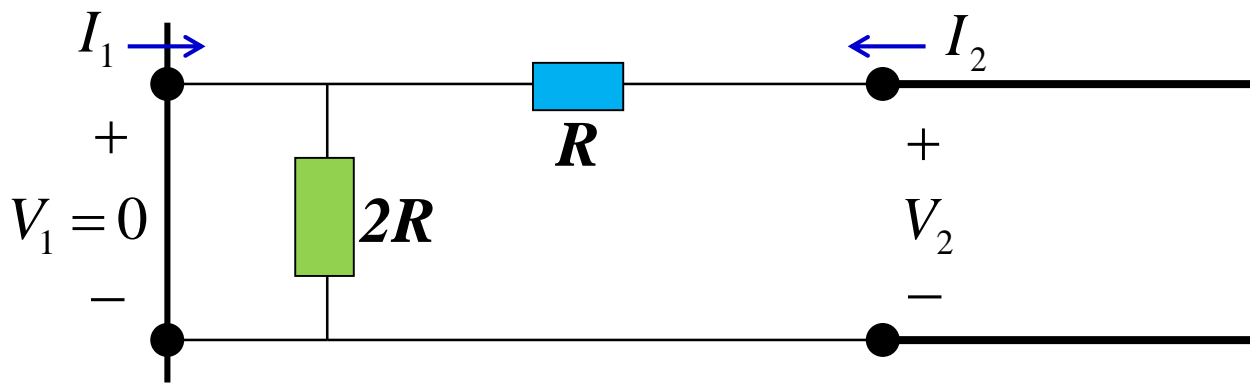
This is **still** a valid physical result, **although** you will find that the **diagonal** terms of an impedance or admittance matrix (e.g., Y_{22} , Z_{11} , Y_{44}) will **always** have a real component that is **positive**

To find the **other two** trans-admittance parameters, we must **move** the short and then **repeat** each of our previous steps!

Example – 1 (contd.)

Step-1:

Place a short at port 1



Step-2:

 Determine currents I_1 and I_2

- Note that **after** a short was placed at port 1, resistor $2R$ has **zero** voltage across it—and thus **zero current** through it!

Therefore:

$$I_2 = \frac{V_2}{R}$$

$$I_1 = -I_2 = -\frac{V_2}{R}$$

Step-3:

Determine the trans-admittances Y_{12} and Y_{22}

$$Y_{12} = \frac{I_1}{V_2} = -\frac{1}{R}$$

$$Y_{22} = \frac{I_2}{V_2} = \frac{1}{R}$$

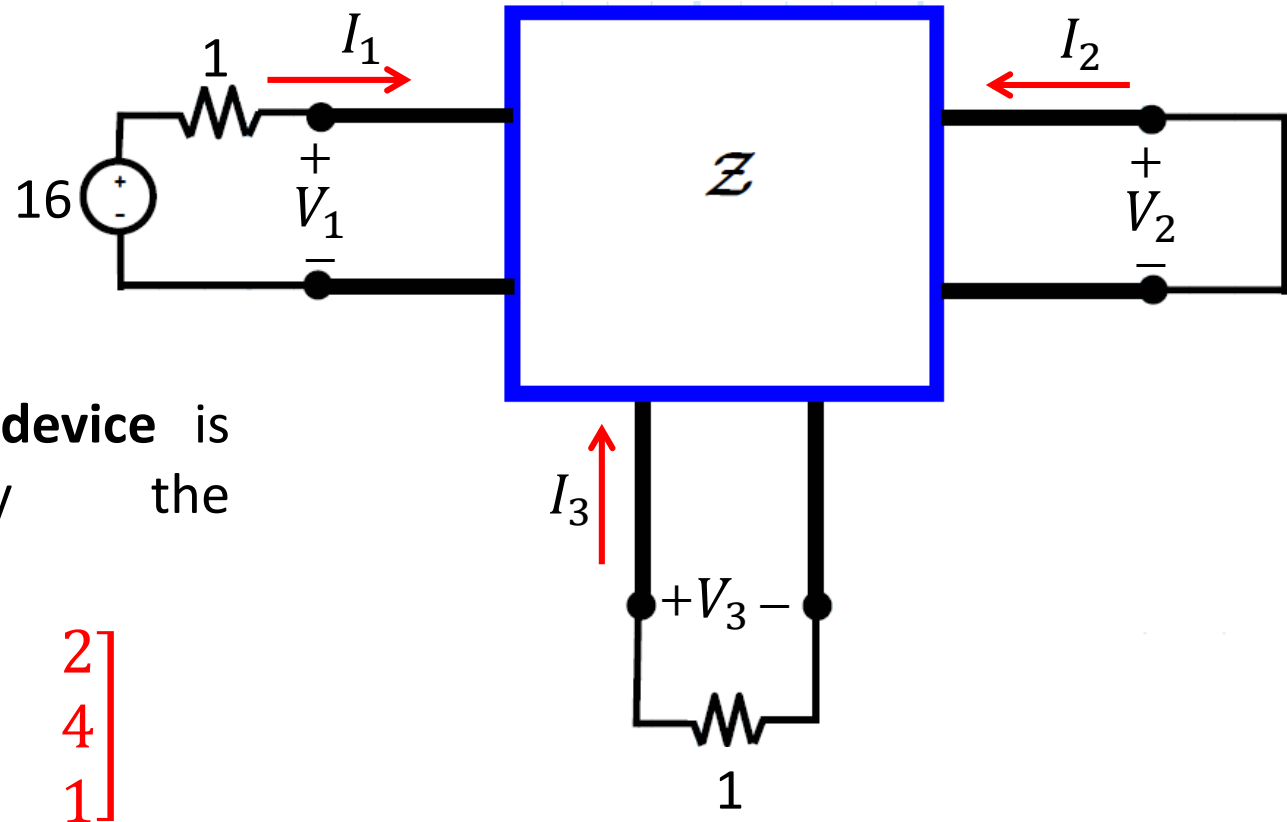
Therefore the admittance matrix is:

$$Y = \begin{bmatrix} 3/2R & -1/R \\ -1/R & 1/R \end{bmatrix}$$

Is it lossless or reciprocal?

Example – 2

- Consider this circuit:



- Where the 3-port **device** is characterized by the **impedance matrix**:

$$Z = \begin{bmatrix} 2 & 1 & 2 \\ 1 & 1 & 4 \\ 2 & 4 & 1 \end{bmatrix}$$

- determine all port **voltages** V_1, V_2, V_3 and all **currents** I_1, I_2, I_3 .