

## Lecture-20

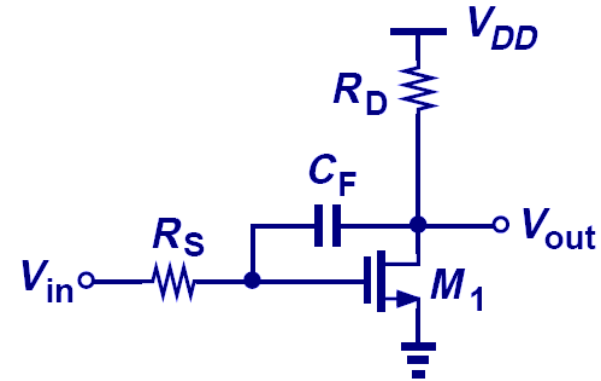
Date: 05.11.2015

- Quiz-7
- Feedback

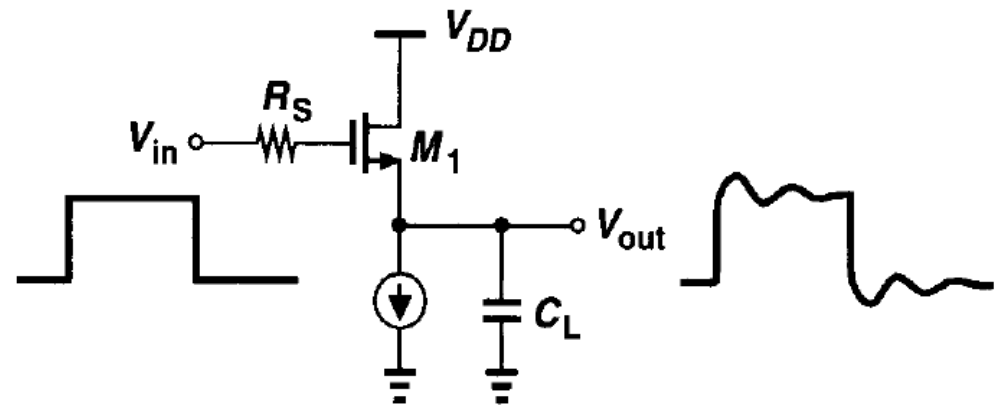
## Quiz – 7

1. What is transit frequency of a MOSFET? What is its significance? **[0.50 marks]**

2. At high frequencies, the frequency response of this amplifier doesn't decay with the usual rule of -20dB/decade? Give justifications. **[1.0 marks]**

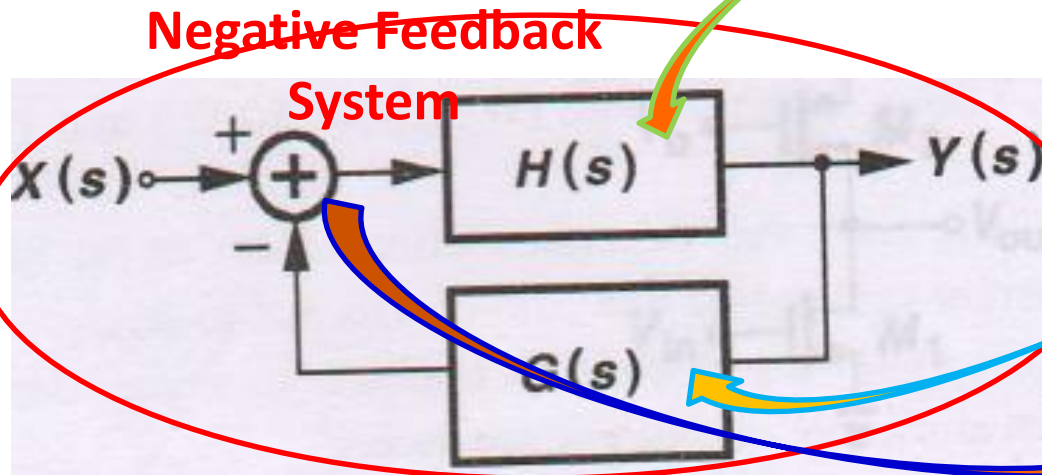


3. What leads to the ringing problem in this amplifier? What could be the solutions to overcome this? **[1.0 marks]**



# Feedback

## General Considerations:



**Negative Feedback System**

**Feedforward Network**

**Feedback Network**

**Subtraction at this node makes this system negative feedback**

$$Y(s) = H(s) [X(s) - G(s)Y(s)]$$

**Generic Equation**

**Open Loop Transfer Function**

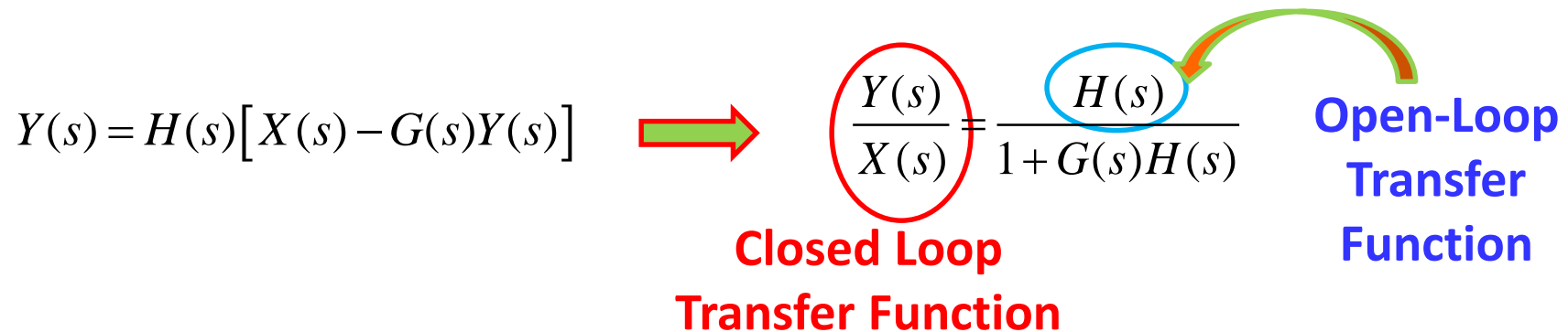
**Error Signal or Compensation Signal**

## Feedback (contd.)

$$Y(s) = H(s)[X(s) - G(s)Y(s)] \quad \longrightarrow \quad \frac{Y(s)}{X(s)} = \frac{H(s)}{1 + G(s)H(s)}$$

Closed Loop  
Transfer Function

Open-Loop  
Transfer  
Function



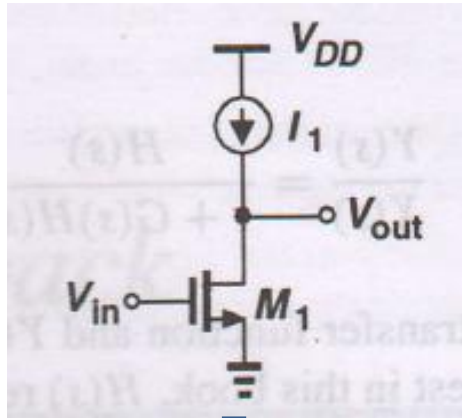
In our discussion: H(s) represents an amplifier and G(s) is a frequency-independent quantity representing the feedback network

### Elements of Feedback System:

- The feed forward amplifier [H(s)]
- A means of sensing the output
- The feedback network [G(s)]
- A means of generating the feedback error [X(s) – G(s)Y(s)]

# Properties of Feedback Circuits

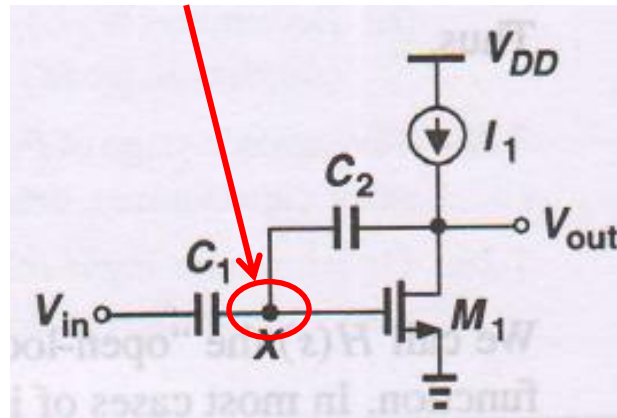
## Gain Desensitization



$$\frac{V_{out}}{V_{in}} = -g_{m1}r_{o1}$$

Dependent on process parameters, temperature, bias etc.

Node generating error term



$$\frac{V_{out}}{V_{in}} = -\frac{1}{\left(1 + \frac{1}{g_{m1}r_{o1}}\right) \frac{C_2}{C_1} + \frac{1}{g_{m1}r_{o1}}}$$

For large

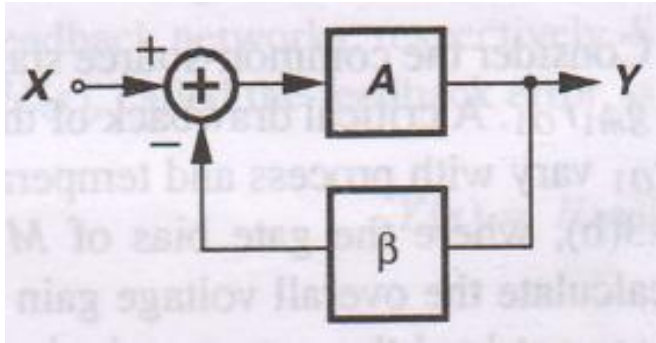
$g_{m1}r_{o1}$

$$\frac{V_{out}}{V_{in}} \approx -\frac{C_1}{C_2}$$

The closed-loop gain is much less sensitive to device parameters

Feedback makes gain of this CS stage independent of process and temperature

## Properties of Feedback Circuits (contd.)



$$\frac{Y}{X} = \frac{A}{1 + A\beta} = \frac{1}{\beta} \frac{1}{\left(1 + \frac{1}{A\beta}\right)}$$

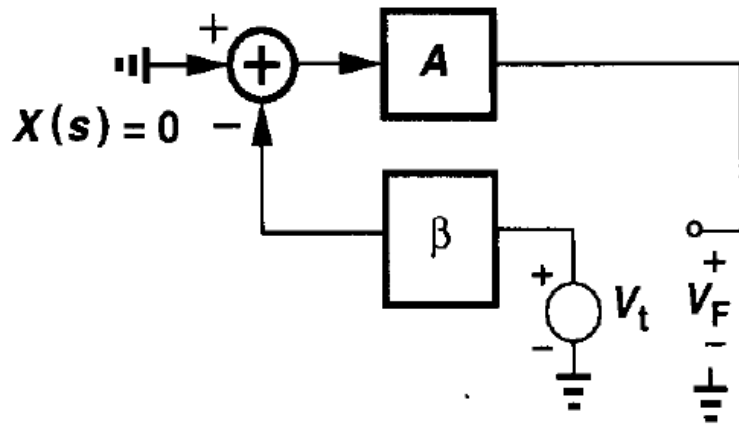
$$\frac{Y}{X} \cong \frac{1}{\beta} \left(1 - \frac{1}{A\beta}\right)$$

The impact of variations in A on the closed loop gain is insignificant

- The quantity  $\beta A$  (loop gain) is critical for any feedback system  $\rightarrow$  higher the  $\beta A$ , less sensitive is the closed loop gain to the variations of A
- Accuracy of the closed loop gain improves by maximizing either A or  $\beta$
- $\beta$  increases  $\rightarrow$  closed loop gain decreases  $\rightarrow$  means a trade-off exist between precision and the closed loop gain

# Properties of Feedback Circuits (contd.)

## Loop Gain Calculation

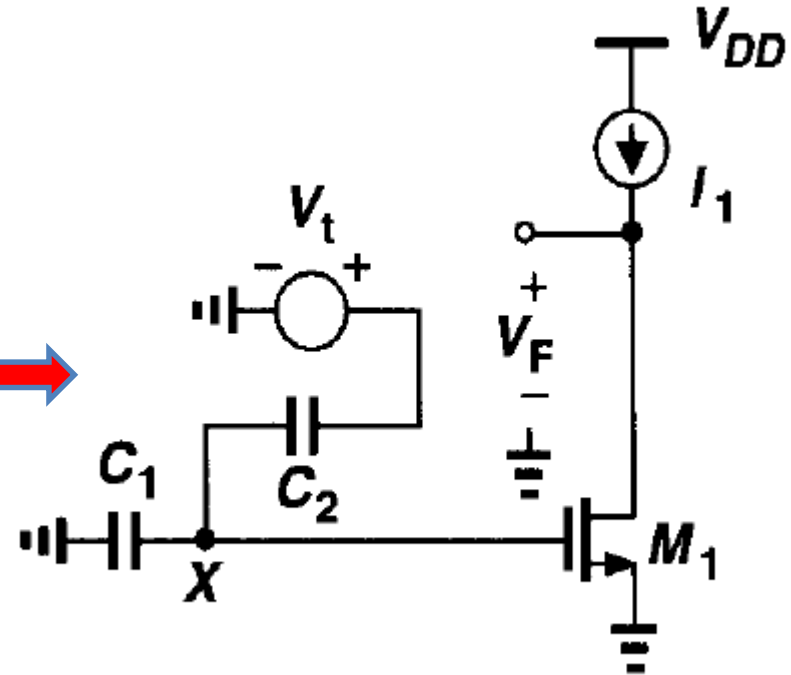
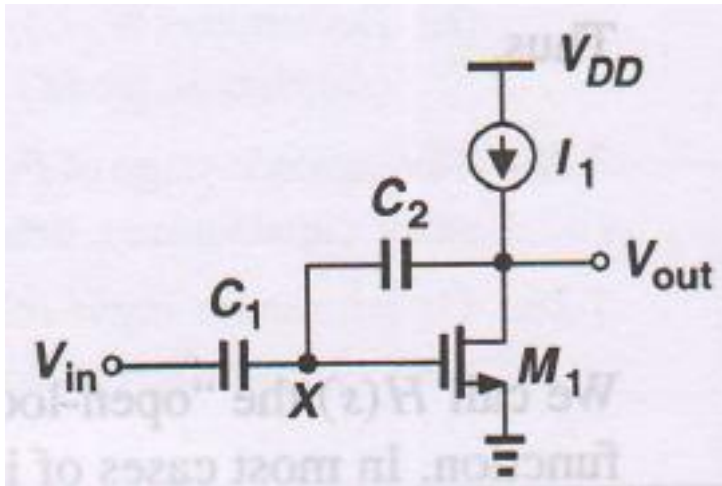


- Set the main input to zero
- Break the loop at some point
- Inject a test signal 'while maintaining the direction and polarity'
- Follow the signal around the loop and obtain the expression/value

$$V_t \beta (-1) A = V_F \quad \longrightarrow \quad \therefore \frac{V_F}{V_t} = -\beta A$$

## Example – 1

Calculate Loop Gain:

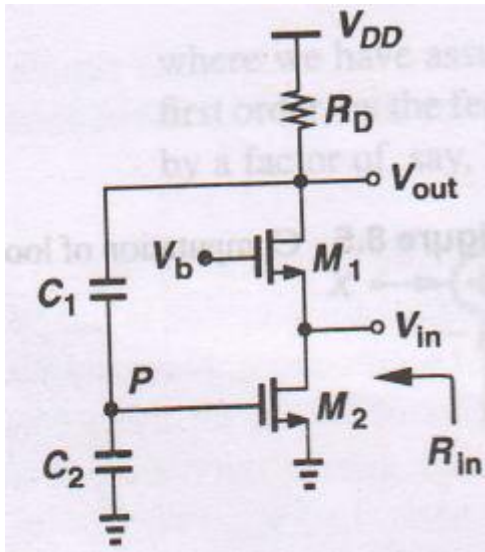


$$\therefore \frac{V_F}{V_t} = -\frac{C_2}{C_1 + C_2} g_{m1} r_{o1}$$

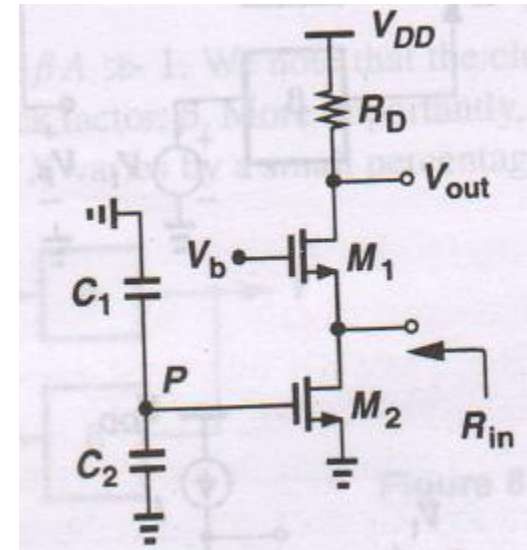


# Properties of Feedback Circuits (contd.)

## Input Impedance Modification



Lets check input  
impedance with  
and without  
feedback



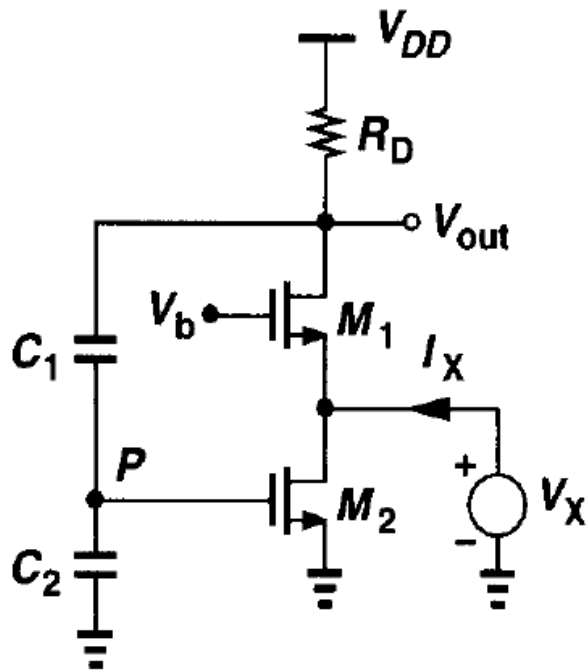
**CG stage ( $M_1$ )** → capacitive divider  
senses  $V_{out}$  and applies it to gate of  
current source ( $M_2$ ) →  $M_2$  returns a  
current feedback signal to the input  
of  $M_1$

$$R_{in,open} = \frac{1}{g_{m1} + g_{mb1}}$$



Assumption: no  
channel-length  
modulation present

## Properties of Feedback Circuits (contd.)



$$\frac{V_X}{I_X} = R_{in,closed} = \frac{1}{g_{m1} + g_{mb1}} \frac{1}{1 + g_{m2} R_D \frac{X_{C1}}{X_{C1} + X_{C2}}}$$

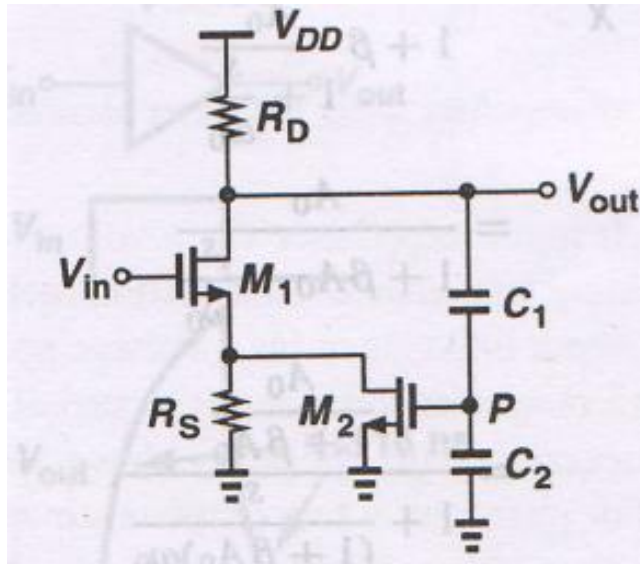
$$R_{in,closed} = R_{in,open} \frac{1}{1 + g_{m2} R_D \frac{X_{C1}}{X_{C1} + X_{C2}}} \text{ Loop Gain}$$

**Feedback reduces the input impedance in this instance → quite useful circuit topology**

**Four Elements of Feedback:** feed-forward amplifier consists of  $M_1$  and  $R_D$ , the output is sensed by  $C_1$  and  $C_2$ , the feedback network comprise of  $C_1$ ,  $C_2$ , and  $M_2$ , subtraction occurs in current domain at the input

## Properties of Feedback Circuits (contd.)

### Output Impedance Modification



$$R_{out,open} = R_D$$

$$R_{out,closed} = \frac{R_D}{1 + \frac{g_{m2} R_S (g_{m1} + g_{mb1}) R_D X_{C1}}{(g_{m1} + g_{mb1}) R_S (X_{C1} + X_{C2})}}$$

**Loop Gain**

**Can you identify if this is a positive feedback or negative feedback circuit? Why?**

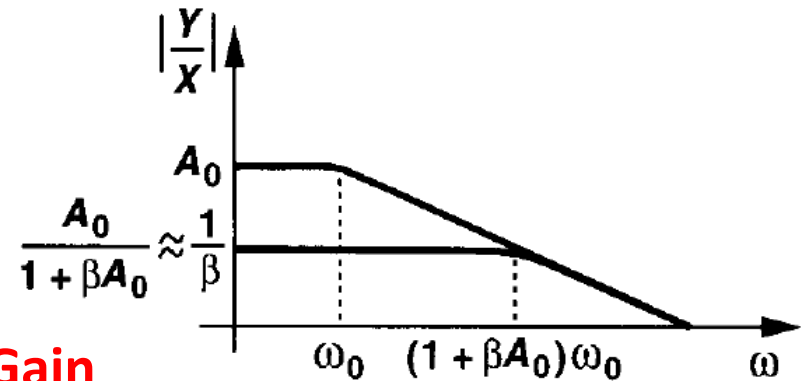
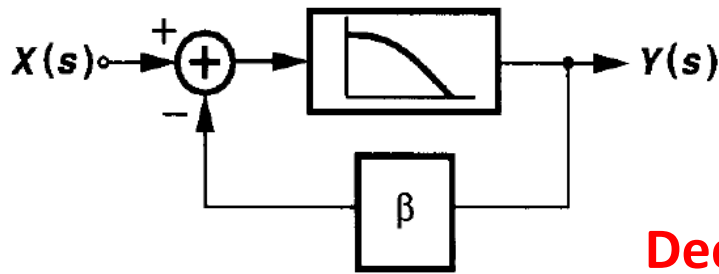
# Properties of Feedback Circuits (contd.)

## Bandwidth Modification

**One pole transfer function:**

$$A(s) = \frac{A_0}{1 + \frac{s}{\omega_0}}$$

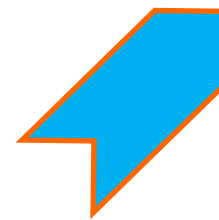
**One pole transfer function of a closed-loop system:**



**Decreased Gain**

$$\frac{Y}{X}(s) = \frac{\frac{A_0}{1 + \frac{s}{\omega_0}}}{1 + \beta \frac{A_0}{1 + \frac{s}{\omega_0}}}$$

$$\frac{Y}{X}(s) = \frac{\frac{A_0}{1 + \beta A_0}}{1 + \frac{s}{(1 + \beta A_0)\omega_0}}$$



**Constant GBW**

**High pole frequency**

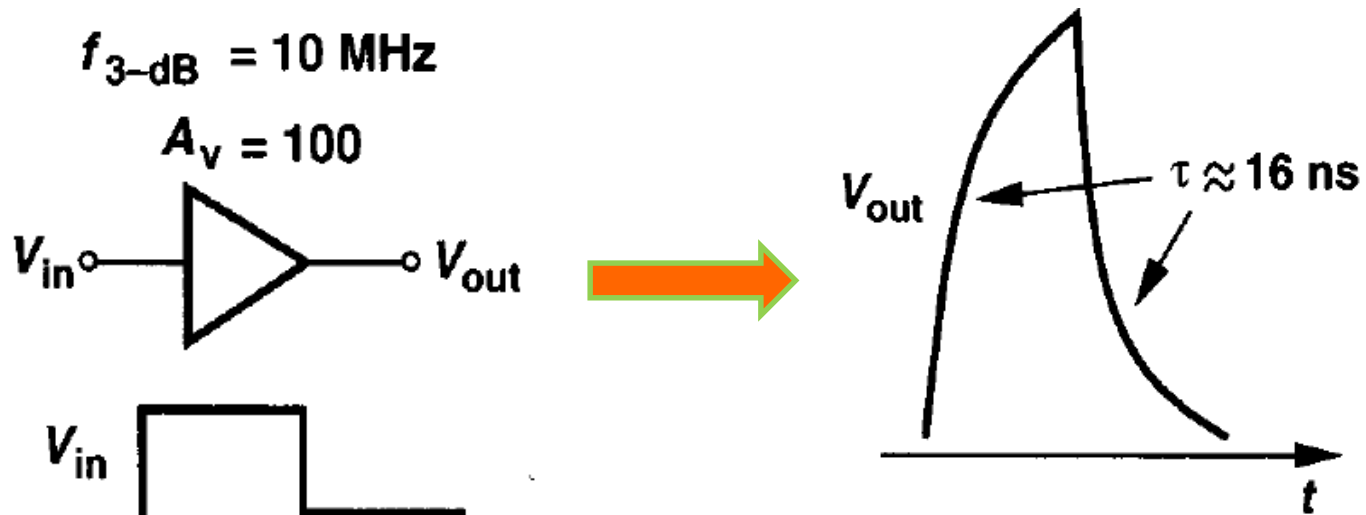
**↔ Increased Bandwidth**

# Properties of Feedback Circuits (contd.)

## Bandwidth Modification

**GBW of a single pole system doesn't change with feedback. But how to improve the speed of the system with high gain?**

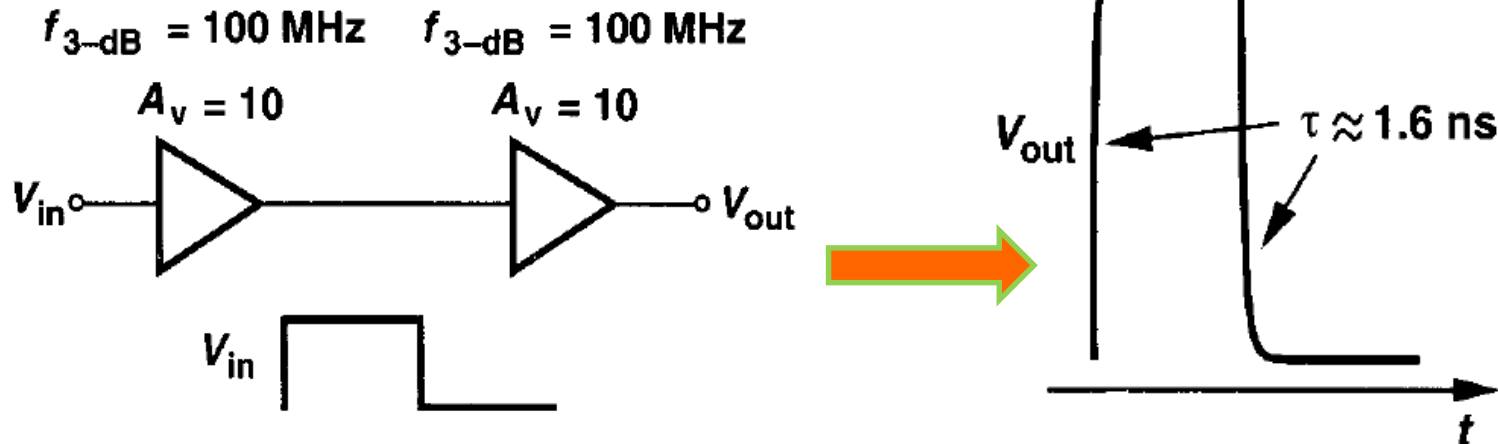
**Suppose we need to amplify a 20 MHz square wave by a factor of 100 and maximum bandwidth but we only have single pole amplifier with an open-loop gain of 100 and 3-dB bandwidth of 10 MHz.**



# Properties of Feedback Circuits (contd.)

## Bandwidth Modification

Apply feedback in such a way that the gain and bandwidth are modified to 10 and 100 MHz. Then use two stage amplification to achieve the desired.



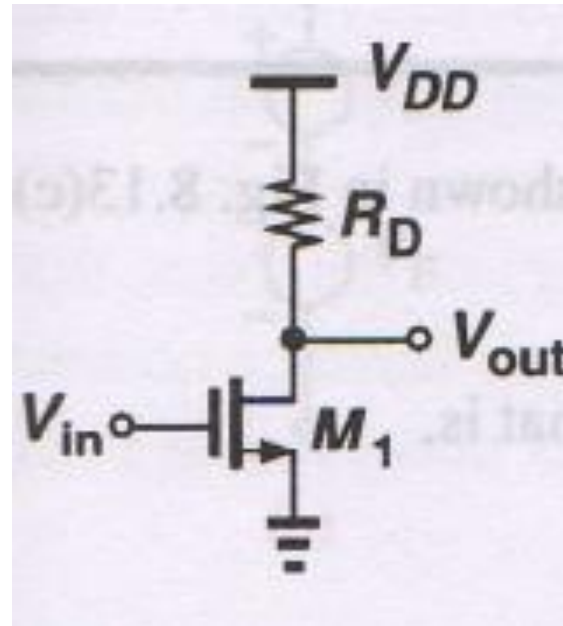
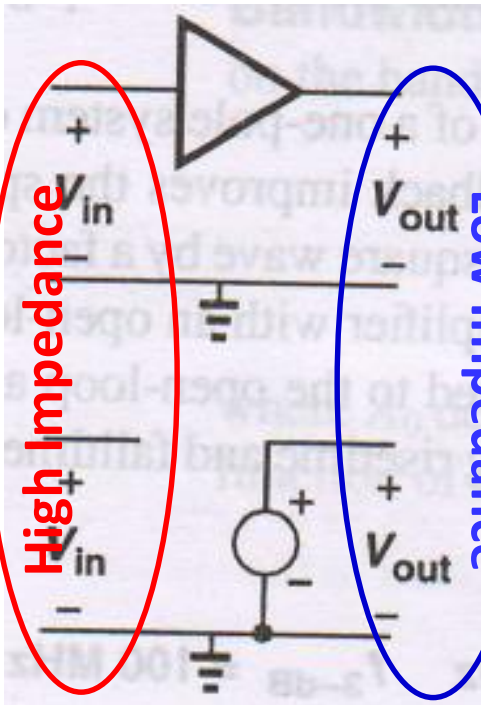
## **Types of Amplifiers**

Type: Based on the type of parameters (current or voltage) they sense at the input and the type of parameters (current or voltage) they produce at the output

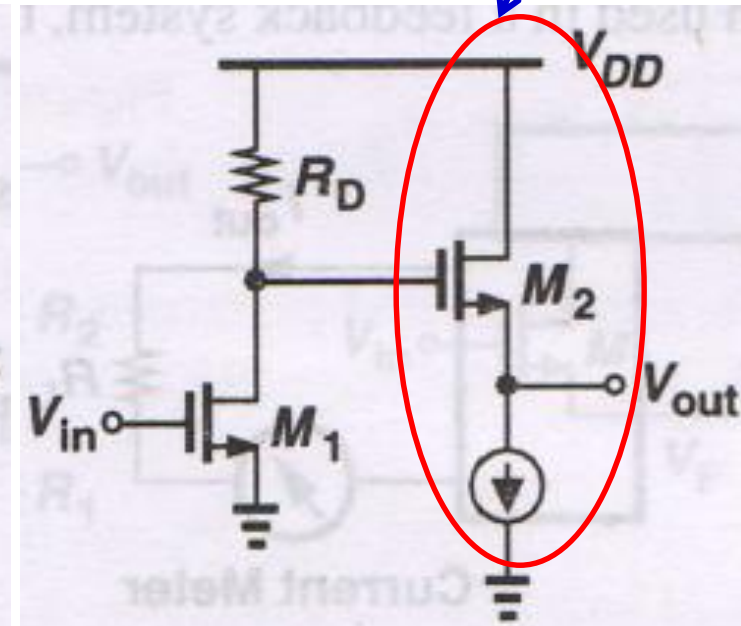
- **Amplifier sensing voltage at the input: exhibit high input impedance (as a voltmeter)**
- **Amplifier sensing current at the input: exhibit low input impedance (as an ammeter)**
- **Amplifier sensing voltage at the output: exhibit low output impedance (as a voltage source)**
- **Amplifier sensing current at the output: exhibit high output impedance (as a current source)**

## Types of Amplifiers (contd.)

### Voltage Amplifier



**CS Amplifier**



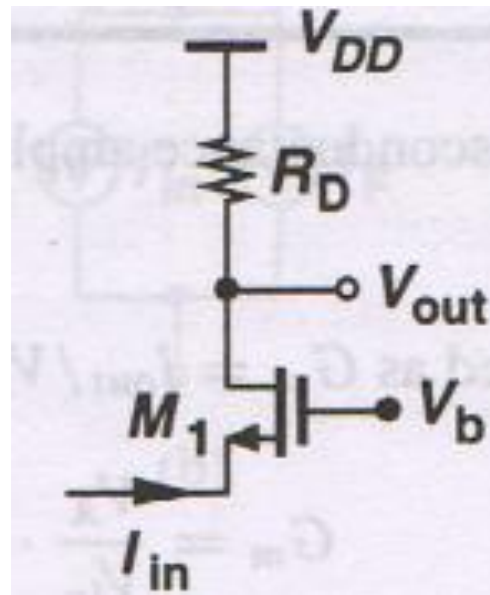
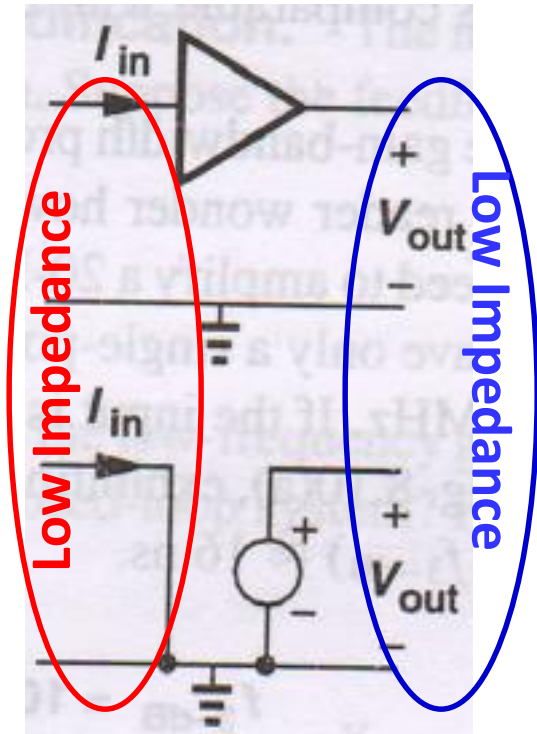
**CS Amplifier with a buffer**

Possess Relatively  
High Output  
Impedance



## Types of Amplifiers (contd.)

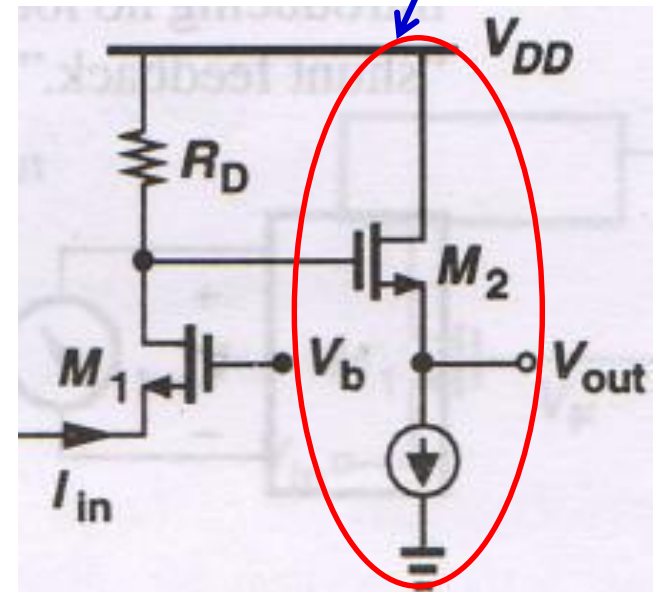
### Transimpedance Amplifier



**CG Amplifier**

Possess Relatively High  
Output Impedance

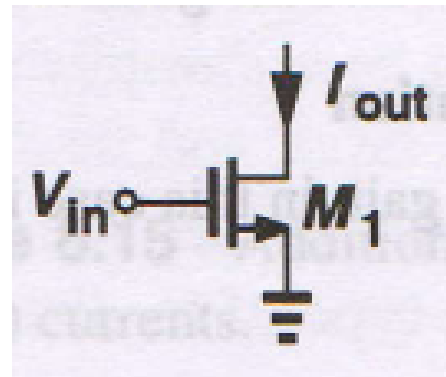
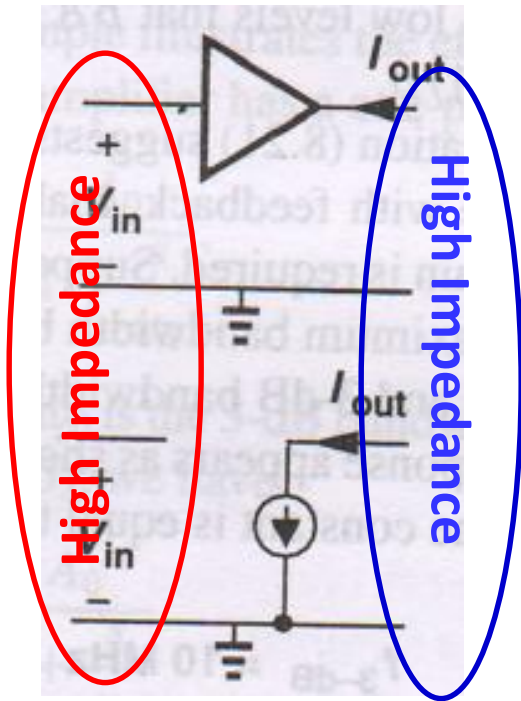
### Buffer / CD Stage



**CG Amplifier with a buffer**

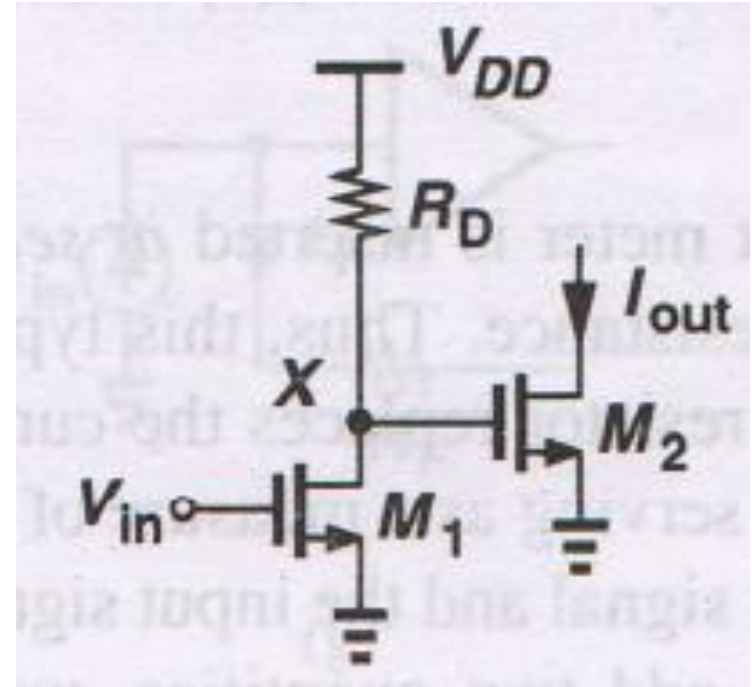
# Types of Amplifiers (contd.)

## Transconductance Amplifier



**CS Stage**

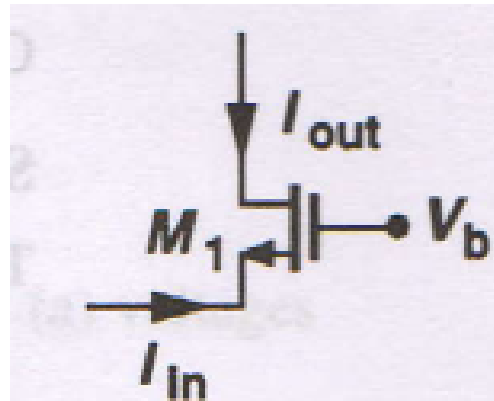
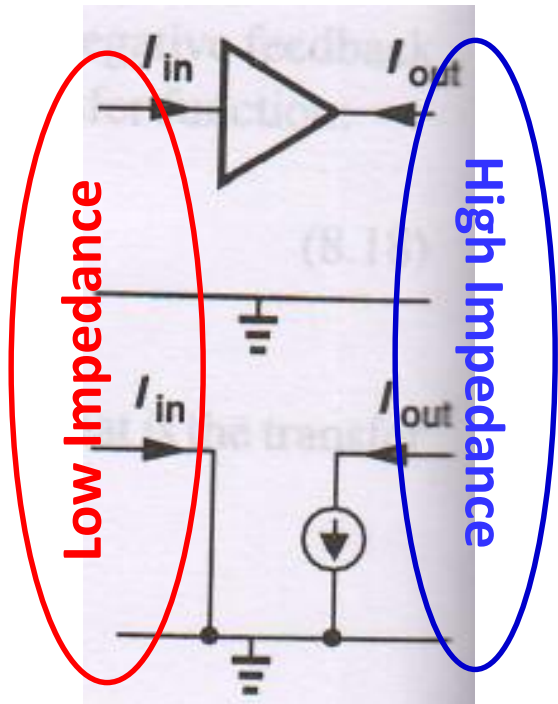
Less control on  
input impedance



**CS Stage with CS  
Amplifier at the input**

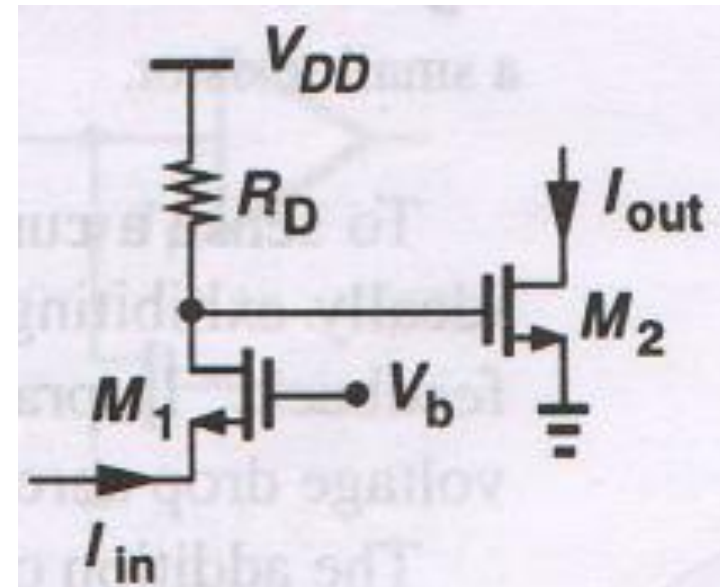
# Types of Amplifiers (contd.)

## Current Amplifier



**CG Stage**

Less control on  
input impedance



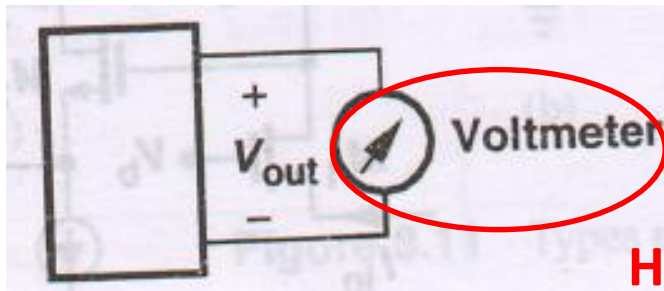
**CG Stage with CG  
Amplifier at the input**

## **Sense and Return Mechanism**

- Placing a circuit in the feedback requires sensing the output signal and then returning a fraction to the input
- Voltage and Current as input and output quantities provide 4 different possibilities for feedback circuit (sense and return circuit)
- **Voltage-Voltage:** both the input and output of the feedback circuit is voltage
- **Voltage-Current:** input of feedback is voltage and output is current
- **Current-Voltage:** input of feedback is current and output is voltage
- **Current-Current:** both the input and output of feedback circuit is current

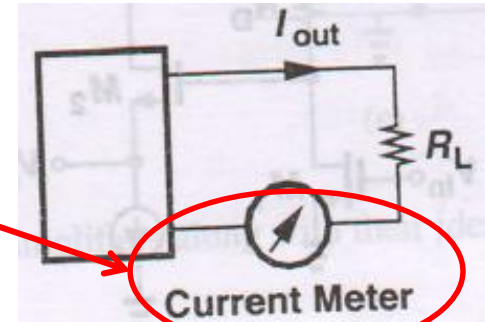
## Sense and Return Mechanism (contd.)

To sense a voltage:



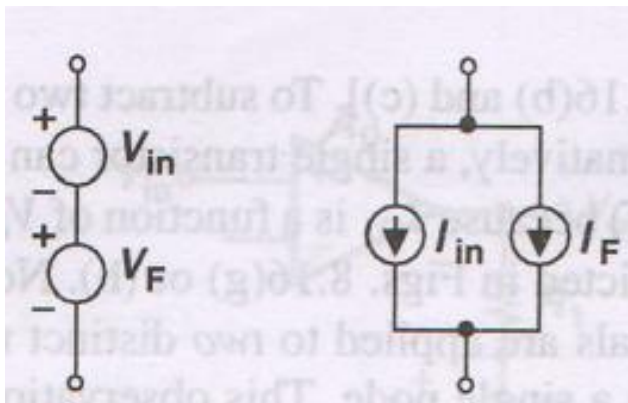
**High Input  
Impedance**

To sense a current:



**Low Input  
Impedance**

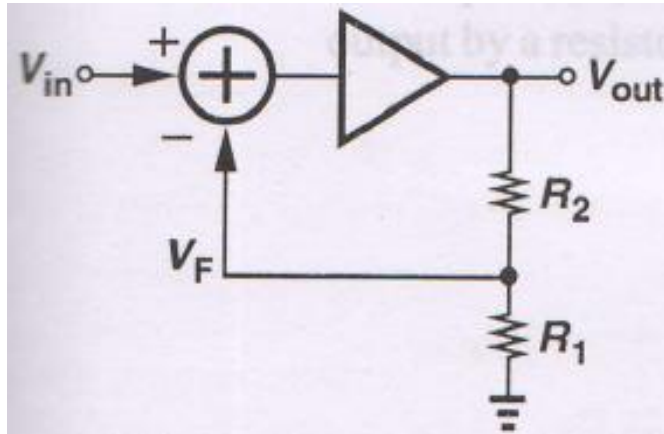
- The addition/subtraction at the input can be done in current or voltage domain: (a) currents are added by placing them in parallel; (b) voltages are added by placing them in series



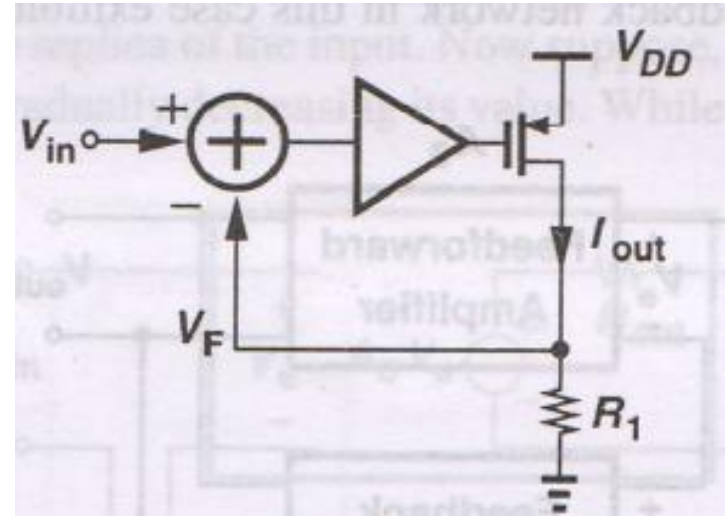
The sense and return mechanism ideally do not affect the operation of feed-forward amplifier → in practical circuits they do introduce loading effects

## Sense and Return Mechanism (contd.)

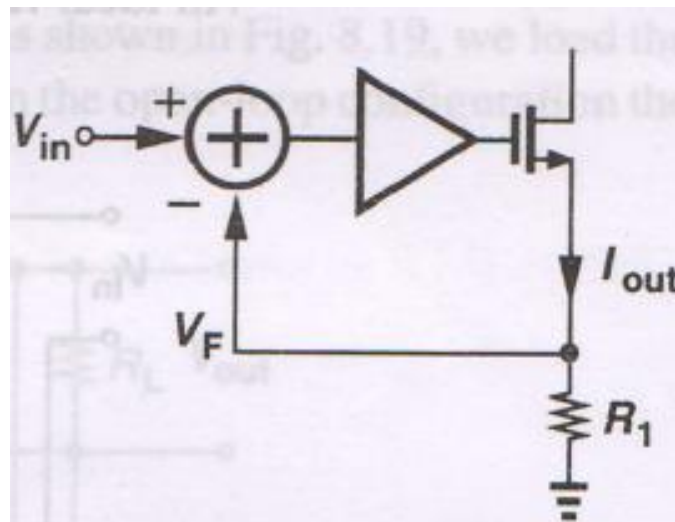
### Practical Implementations of Sensing:



**Voltage Sensing**



**Current Sensing**

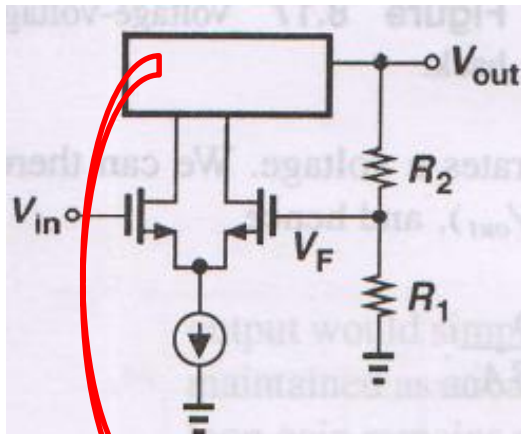


**Current Sensing**

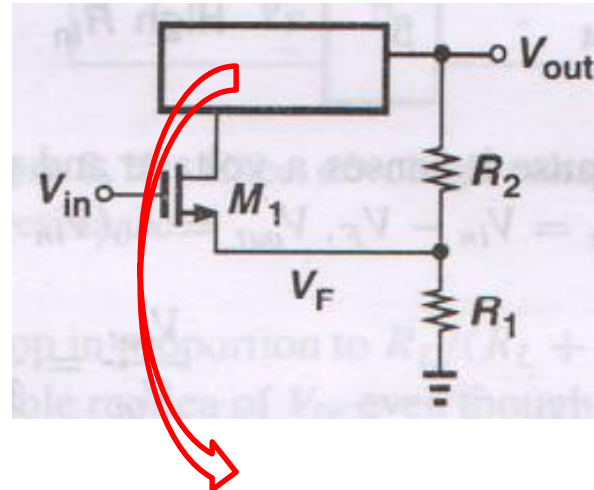


## Sense and Return Mechanism (contd.)

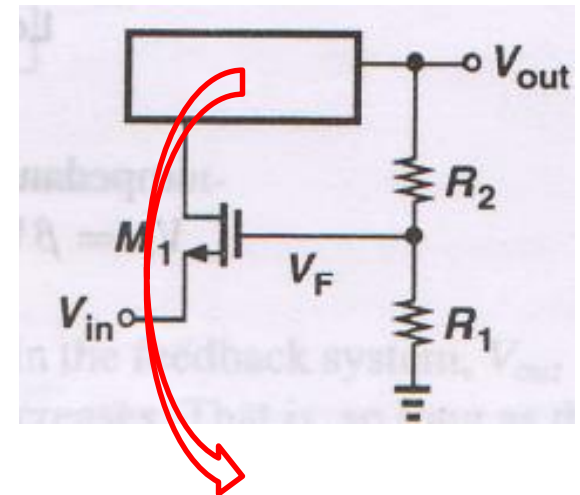
### Practical Implementations of Voltage Subtraction:



Provides the amplified version of difference between  $V_{in}$  and the portion of  $V_{out}$



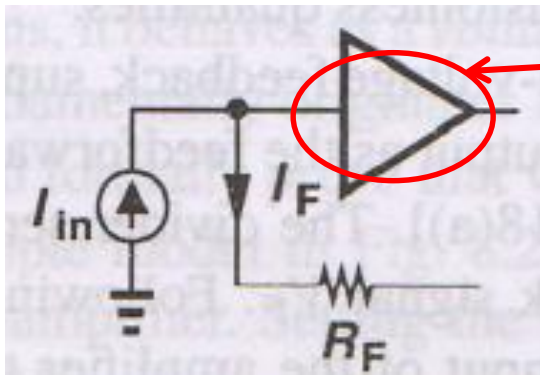
This CS stage provides output in terms of voltage difference  $V_{in} - V_F$



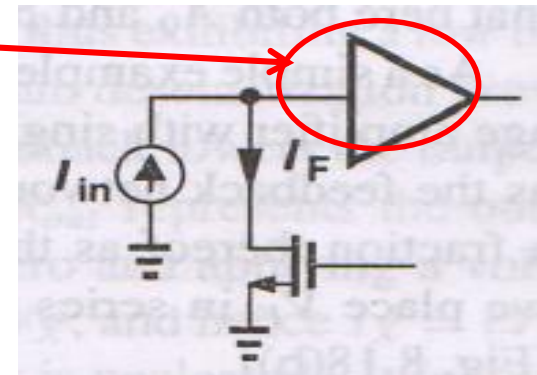
This CG stage provides output in terms of voltage difference  $V_{in} - V_F$

## Sense and Return Mechanism (contd.)

### Practical Implementations of Current Subtraction:



**Feed-forward  
Amplifier**



**Important:** voltage subtraction happens when they are applied to two distinct nodes whereas current subtraction happens when they are applied to a single node → **a precursor to feedback topologies**